

ASET COLLOQUIUM / NSPDI 2017
Tata Institute of Fundamental Research, Mumbai

High-speed data communication links for high energy physics experiments

Shalabh Gupta (shalabh@ee.iitb.ac.in)
Associate Professor
Department of Electrical Engineering, IIT Bombay
Powai, Mumbai – 400076, INDIA

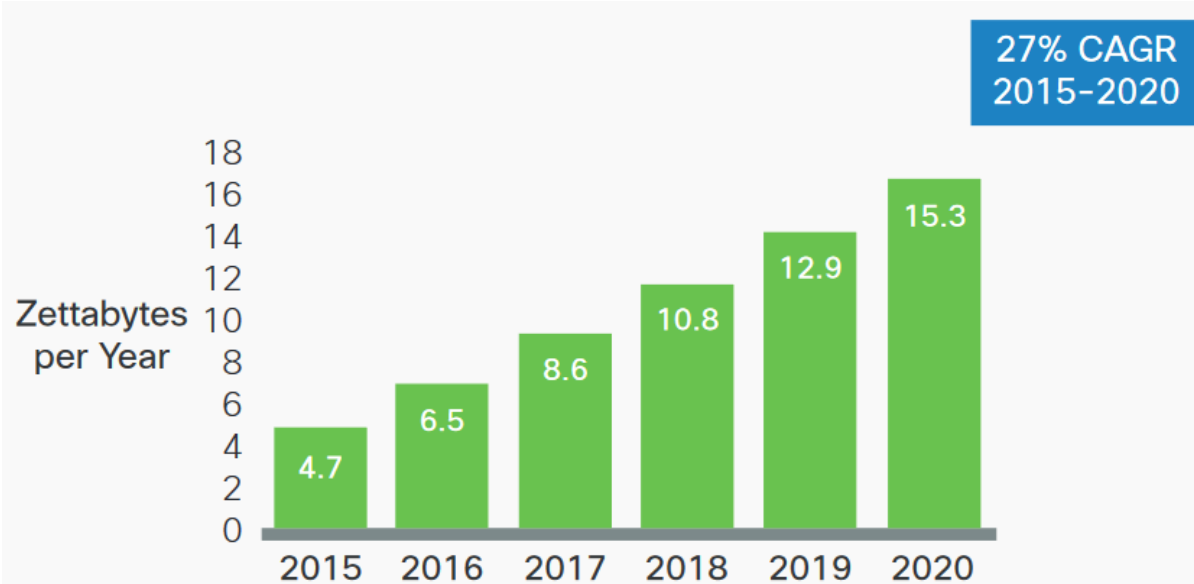
October 6, 2017



Need for High-Speed Communication Links

Explosive growth in the Internet traffic

Global Data Center IP Traffic Forecast



Source: Cisco Global Cloud Index, 2015-2020.

- **Solution: Use of coherent modulation and detection: Electronics – the key enabler!**
- **Optical links with 200 – 400 Gbps/wavelength** being developed and deployed for long haul.



Recent Focus: Data Centers

News & Analysis

Facebook Likes 100G at \$1/G

Rick Merritt

8/28/2015 11:50 AM EDT
2 comments

Tweet 11 Share 78 G+ 5

Web giant plans 2016 Ethernet upgrade

SANTA CLARA, Calif. – Facebook helped define a low cost 100G transceiver it expects to start deploying next year. The move at data centers are driving design in computers and networking.

Facebook specified a 100G transceiver using single-mode fiber. **To hit the lower costs it relaxed distance requirements to 500 meters down from 2km and eased specs on operating temperature and product lifetime.**

Inphi Debuts 100G DWDM Solution for 80km Data Center Interconnects

ColorZ™ First to Enable 4Tb/s Bandwidth in QSFP28 Form Factor while Dramatically Lowering Power and Cost

SANTA CLARA, Calif., March 22, 2016- Inphi Corporation (NYSE: INPH) provider of high-speed connectivity solutions, today announced the release of its "ColorZ" reference design, the industry's first Silicon Photonics 100G PAM4 platform solution for 80km DWDM Data Center Interconnect (DCI) in QSFP28 form factor.

Utilizing advanced Pulse Amplitude Modulation (PAM4) signaling, ColorZ delivers up to 4Tb/s of bandwidth over a single fiber and allows multiple data centers located up to 80km of each other to be connected and act like a single data center. ColorZ provides

OFC 2016

OFC
The Optical Networking and Communication Conference & Exhibition
Technical Conference: 11 - 15 March 2018
Exhibition: 13 - 15 March 2018
San Jose, California, USA

ABOUT PROGRAM & SPEAKERS EXHIBIT HALL & FLOOR PROGRAMS SUBMIT PAPERS

HOME > PROGRAM & SPEAKERS > 2017 PANELS

Direct vs. Coherent Detection for Metro-DCI

Access Technical Papers
2017 Plenary
2017 Invited Speakers
2017 Tutorial Speakers
2017 Data Center Summit
2017 Demonstrations and Events

Tuesday, March 21, 2017
4:30 PM - 6:30 PM
Event type: Panel
Room number: 402AB
Organizer:
Robert Griffin, Oclaro, UK; Tom Issenhuth, Microsoft, USA; Sorin Tibuleac, Adva Optical, USA

OFC 2017 Panel

High-capacity long haul links have been there for a while. Focus now shifting to short distance Tbps optical interconnects

CMS Detector @CERN

CMS DETECTOR

Total weight : 14,000 tonnes
 Overall diameter : 15.0 m
 Overall length : 28.7 m
 Magnetic field : 3.8 T

STEEL RETURN YOKE
 12,500 tonnes

SILICON TRACKERS
 Pixel (100x150 μm) $\sim 16\text{m}^2 \sim 66\text{M}$ channels
 Microstrips (80x180 μm) $\sim 200\text{m}^2 \sim 9.6\text{M}$ channels

SUPERCONDUCTING SOLENOID
 Niobium titanium coil carrying $\sim 18,000\text{A}$

MUON CHAMBERS
 Barrel: 250 Drift Tube, 480 Resistive Plate Chambers
 Endcaps: 468 Cathode Strip, 432 Resistive Plate Chambers

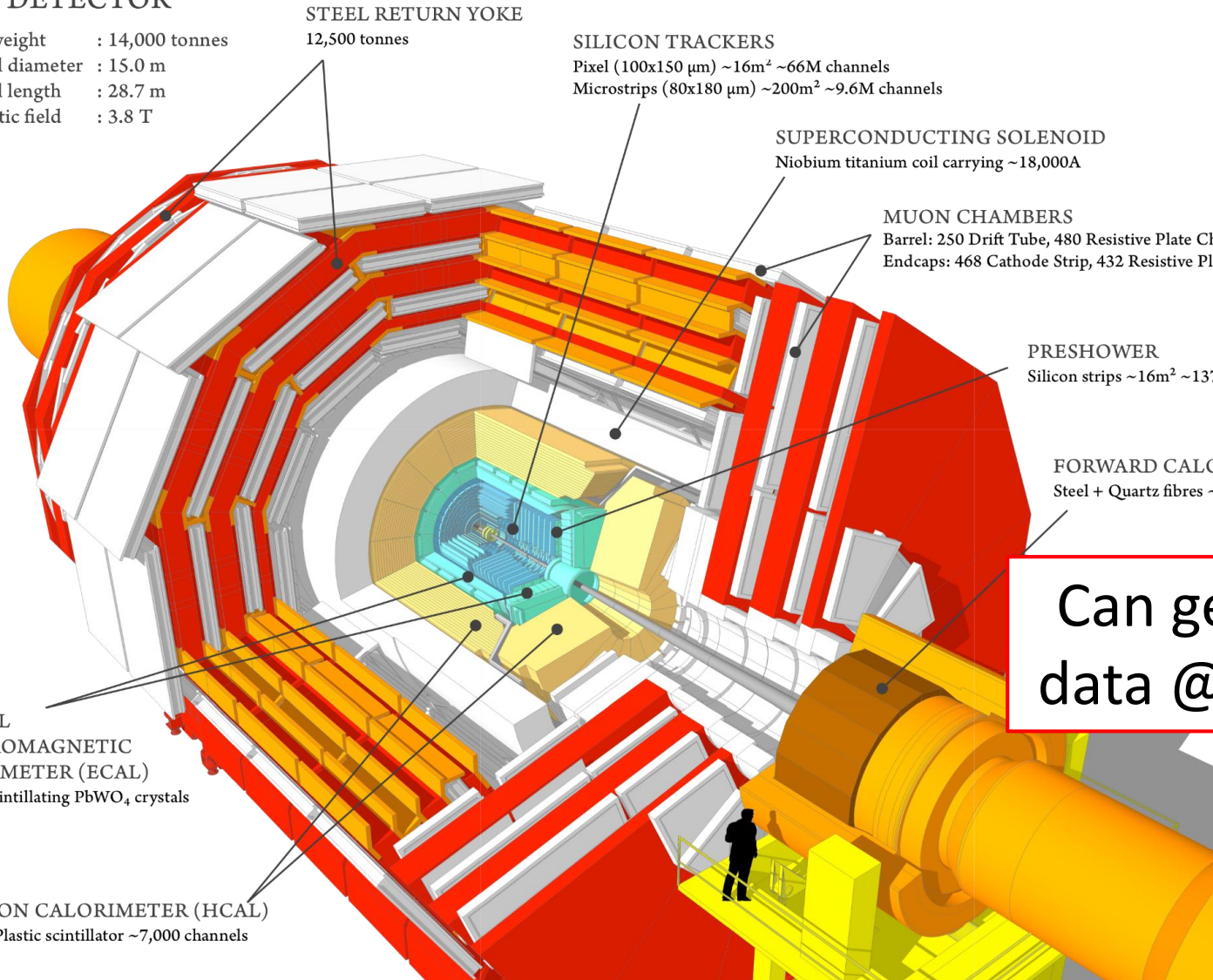
PRESHOWER
 Silicon strips $\sim 16\text{m}^2 \sim 137,000$ channels

FORWARD CALORIMETER
 Steel + Quartz fibres $\sim 2,000$ Channels

CRYSTAL
 ELECTROMAGNETIC
 CALORIMETER (ECAL)
 $\sim 76,000$ scintillating PbWO_4 crystals

HADRON CALORIMETER (HCAL)
 Brass + Plastic scintillator $\sim 7,000$ channels

Can generate
 data @40 Tb/s





Outline

- High-speed serial links: Basic overview
 - Clocking / Clock and Data Recovery
 - Serializers & Deserializers
 - Current progress in high-speed serial links
 - Considerations – for HEP experiments
- Optical communication links: An overview
- Current trends: Digital Coherent Transmission Systems
- Research at IIT Bombay: Low power coherent optical links



Disclaimer: Have no exposure to the HEP experiments myself!



Problem Statement

- Millions of sensor channels produce tens of Terabits/second of data
- Data from various channels has to be aggregated and transferred to central/remote locations for processing/storage
- Latency has to be minimized for triggers and control
- Power efficiency may also be important (adversely affect the temperatures in the detector)



Overview of high-speed serial links

High-Speed Interfaces

Parallel Interfaces for Higher Speeds

- Use many interconnects in parallel
- Crosstalk, mismatches still limit the speed
- Examples:
 - PCI – Peripheral Component Interface
 - Upto 533 MB/s (64-bit, 66 MHz): Has become obsolete
 - GPIB – General Purpose Interface Bus
 - Upto only a few MB/s (8-bit parallel data)
 - Many test equipment still have it (although very inconvenient to use)



Parallel ATA (for HDDs)
– up to 133 MB/s





High-Speed Serial Interfaces

High-Speed Serial Interfaces

- Smaller form factor / area on PCB
 - Faster speeds
 - Lower power
 - More reliable
-
- Examples
 - USB (v1.0: 1.5 Mbit/s to v3.1: 10 Gbit/s)
 - HDMI (v1.0: 1.65 Gbps/lane to v2.1: 12 Gbps/lane)
 - PCIe: PCI-Express (v1.0: 2.5 Gbps/lane to v4.0: 16 Gbps/lane, upto 16 lanes)
 - SATA (upto 16 Gbps) for HDD access





High-Speed Serial Interfaces: Key Enablers

- Low-voltage differential signaling
 - Common-mode noise is suppressed
 - Results in lower power consumption
- Availability of fast, low power, low cost electronics
 - Clock recovery/alignment techniques
 - Equalization techniques
 - Line coding techniques
 - Error correction coding/decoding
- Necessity!!

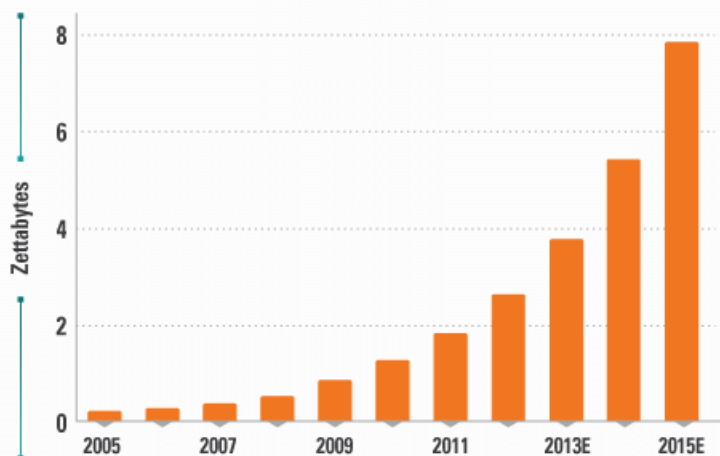


Need for High-Speed Serial Links

- Higher speed links: Some more examples
 - Communication between ADC/DACs and FPGAs/DSP
 - Example: A quad of 16-bits 1-GS/s DAC (used in Mobile base-stations) – requires 64 Gbps delivered to the chip
 - Automotive industry – Self driven cars
 - Satellites: Communications, Imagery data
 - Data Centers – Working as Clouds

Xilinx FPGAs Available (Feb. '17)

A Digital Data Explosion
Global digital information created and shared



Source: KPCB, IDC

techandinnovationdaily.com

| | Type | Max Performance (Gbps) | Max Transceivers | Peak Bandwidth (bidirectional) |
|--------------------|-------------|------------------------|-----------------------|--------------------------------|
| Virtex UltraScale+ | GTY | 32.75 | 128 | <u>8,384 Gb/s</u> |
| Kintex UltraScale+ | GTH/GTY | 16.3/32.75 | 44/32 | 3,268 Gb/s |
| Virtex UltraScale | GTH/GTY | 16.3/30.5 | 60/60 | 5,616 Gb/s |
| Kintex UltraScale | GTH | 16.3 | 64 | 2,086 Gb/s |
| Virtex-7 | GTX/GTH/GTZ | 12.5/13.1/28.05 | 56/96/16 ³ | 2,784 Gb/s |
| Kintex-7 | GTX | 12.5 | 32 | 800 Gb/s |
| Artix-7 | GTP | 6.6 | 16 | 211 Gb/s |
| Zynq UltraScale+ | GTR/GTH/GTY | 6.0/16.3/32.75 | 4/44/28 | 3,268 Gb/s |
| Zynq-7000 | GTX | 12.5 | 16 | 400 Gb/s |
| Spartan-6 | GTP | 3.2 | 8 | 51 Gb/s |

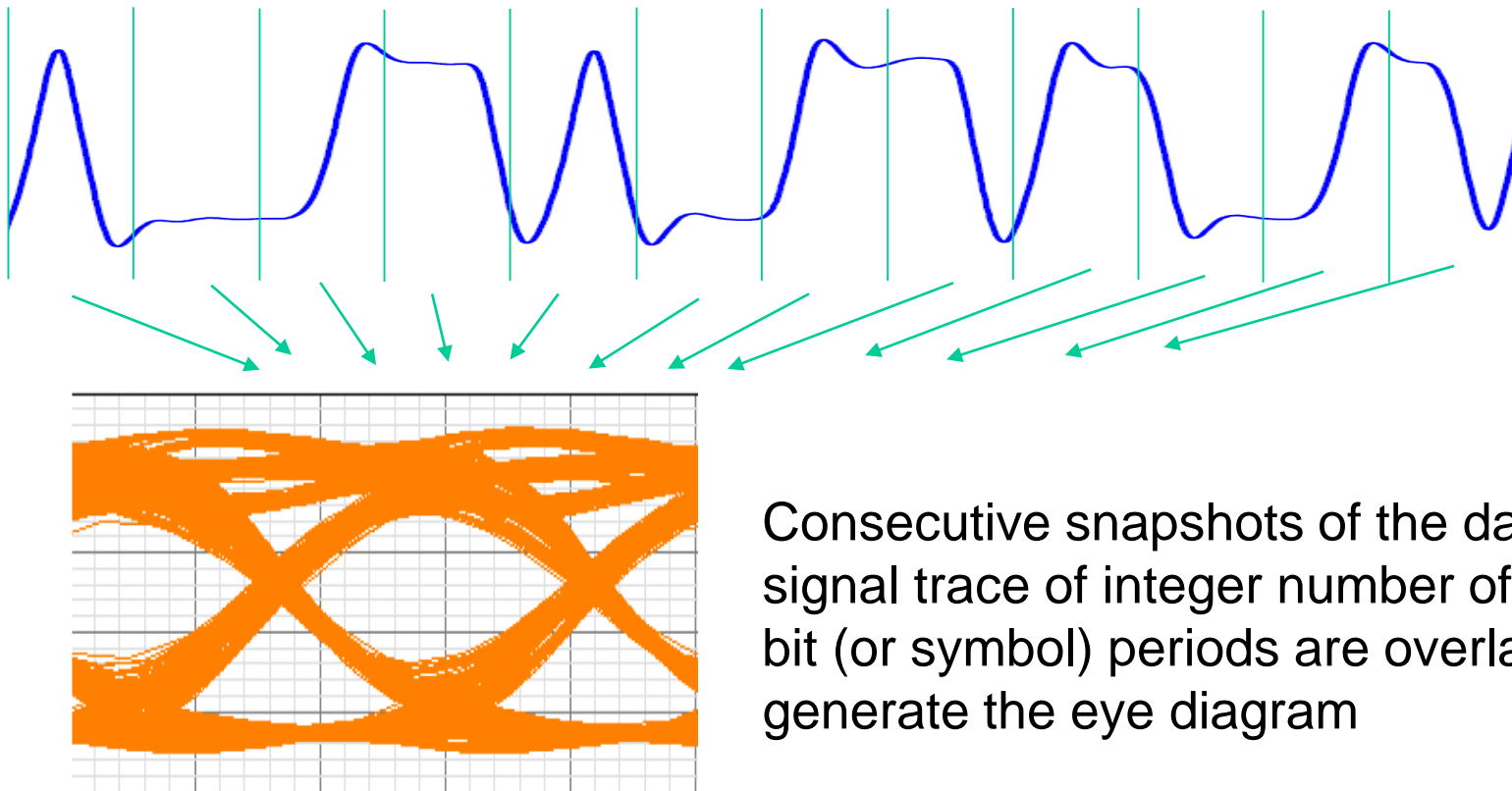


Challenges in Serial Links

- Digital data signals no longer appear to have well defined HIGH or LOW levels.
- There is significant ISI (Inter-Symbol-Interference) due to
 - Impedance mismatches
 - Limited bandwidth of the channel
- Due to ISI, poor eye opening is obtained
 - High BER or no signal detection at all!

Eye Diagrams

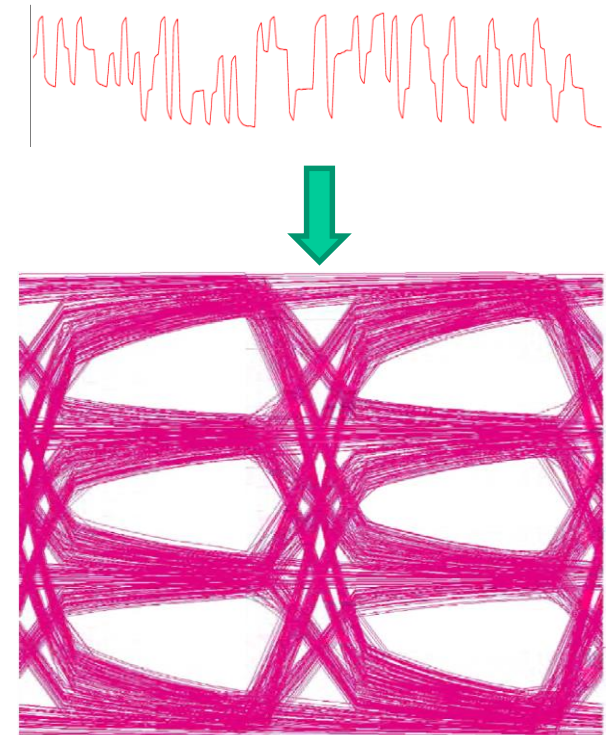
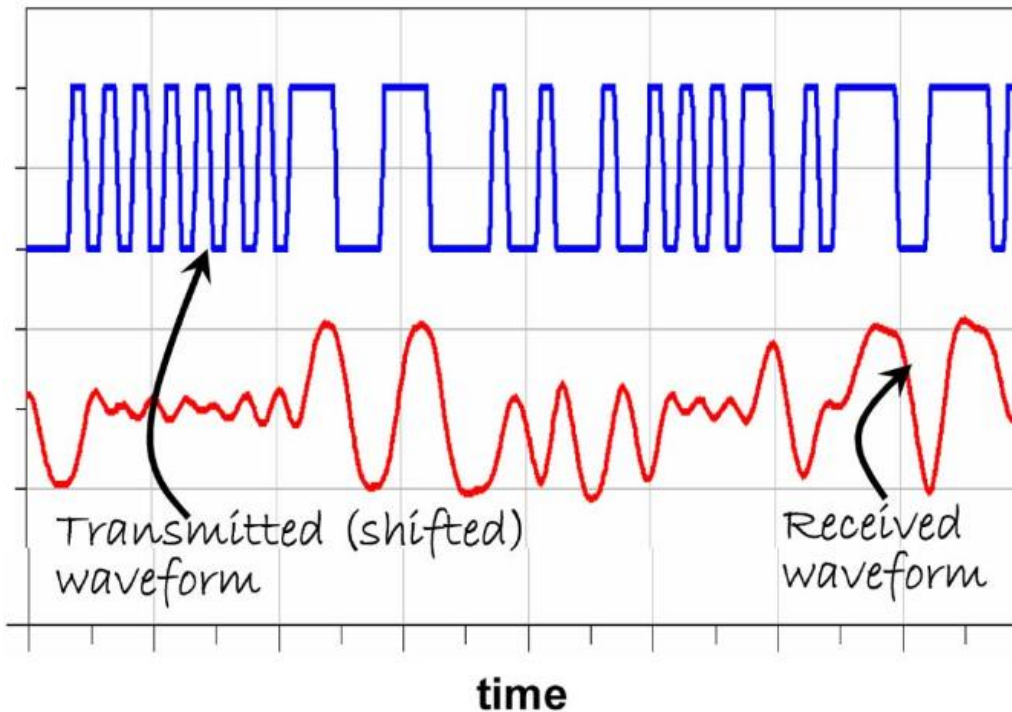
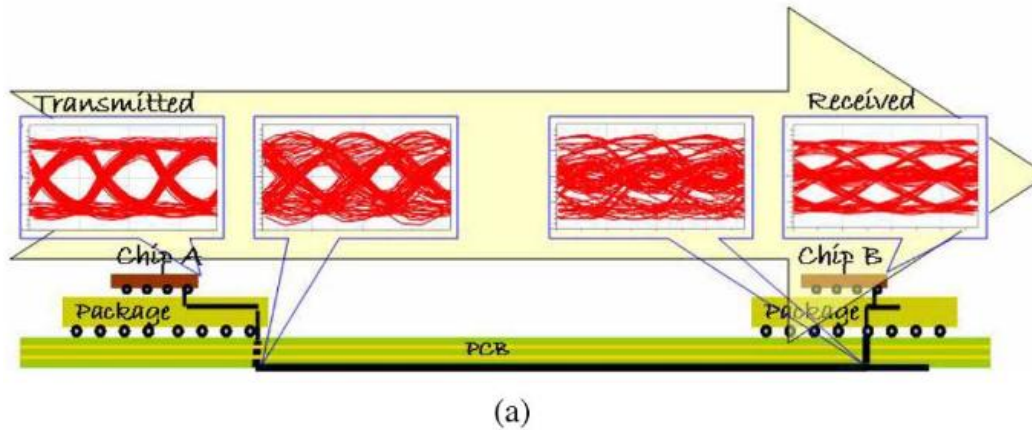
- Eye diagram is used for visualizing data pattern that is synchronous with clock source
- Periodic snapshots are overlaid in the oscilloscope/display to obtain the eye diagram



Consecutive snapshots of the data signal trace of integer number of data bit (or symbol) periods are overlaid to generate the eye diagram

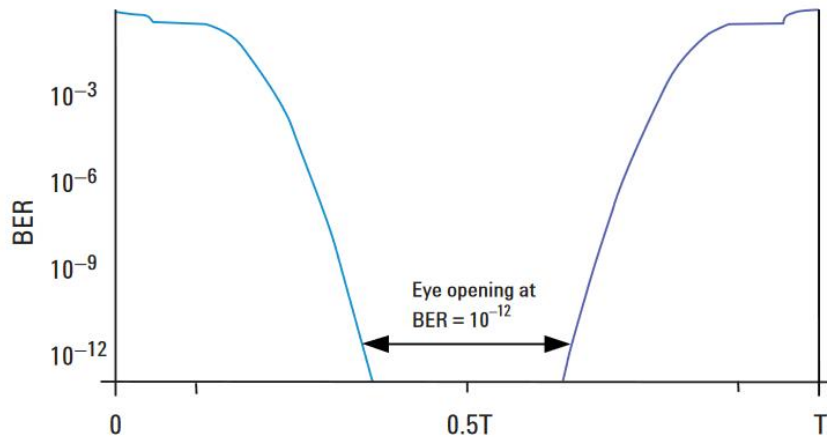
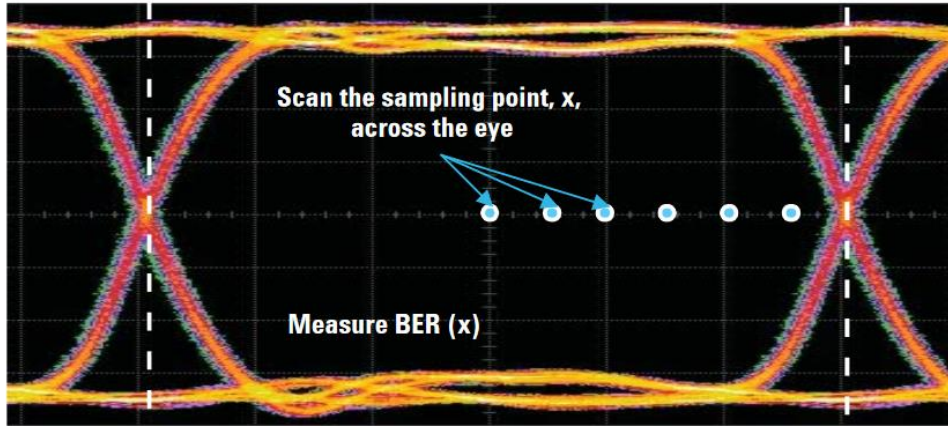
Eye Diagrams: Example

[Beyene, IEEE TAP, 2008]



Eye diagram for PAM-4

Eye Diagrams: BER Estimation



$$TJ (BER) = T - (\text{Eye opening at BER})$$

Noise and Jitter can be determined from the eye-diagrams to estimate Bit-Error-Ratio (BER)

[Agilent Technologies Application Note: Using Clock Jitter Analysis to Reduce BER in Serial Data Applications]



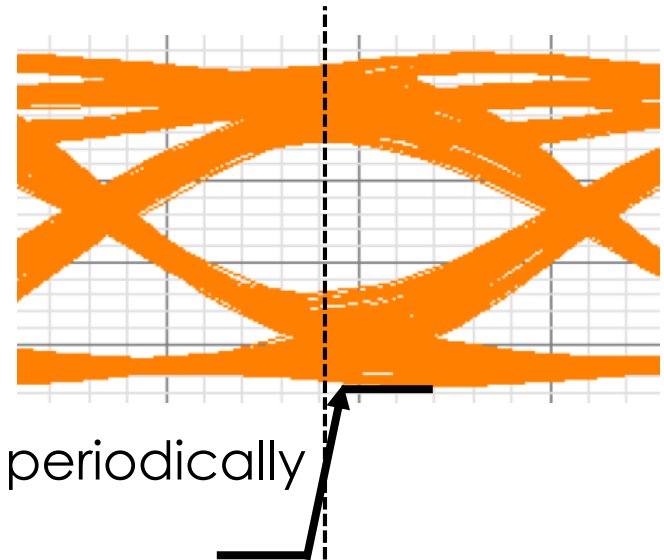
Clocking in High-Speed Serial Interfaces

Clocking in Electronic Systems

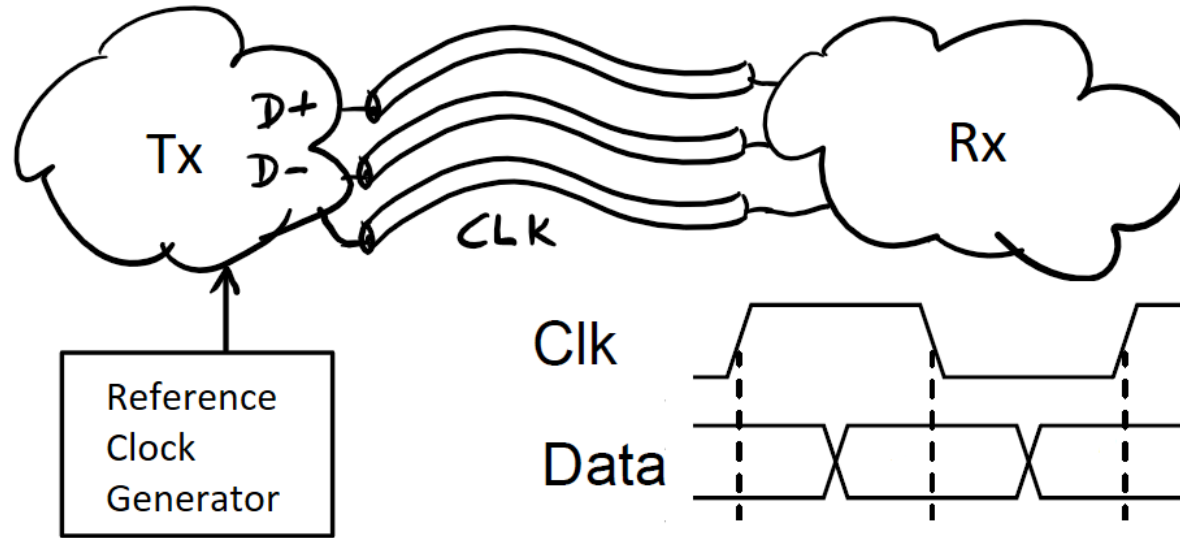
- Clocks in ICs are generally derived from various reference oscillators or incoming reference clocks.
- Independent PCBs typically have have independent clock references
- Based on clocking, systems can be classified as
 - Synchronous Systems
 - Mesochronous Systems
 - Plesiochronous
 - Asynchronous

Synchronous/Mesochronous systems are Isochronous (having same frequency)

Sampling edge moves across the data eye periodically for a fixed sampling clock frequency offset

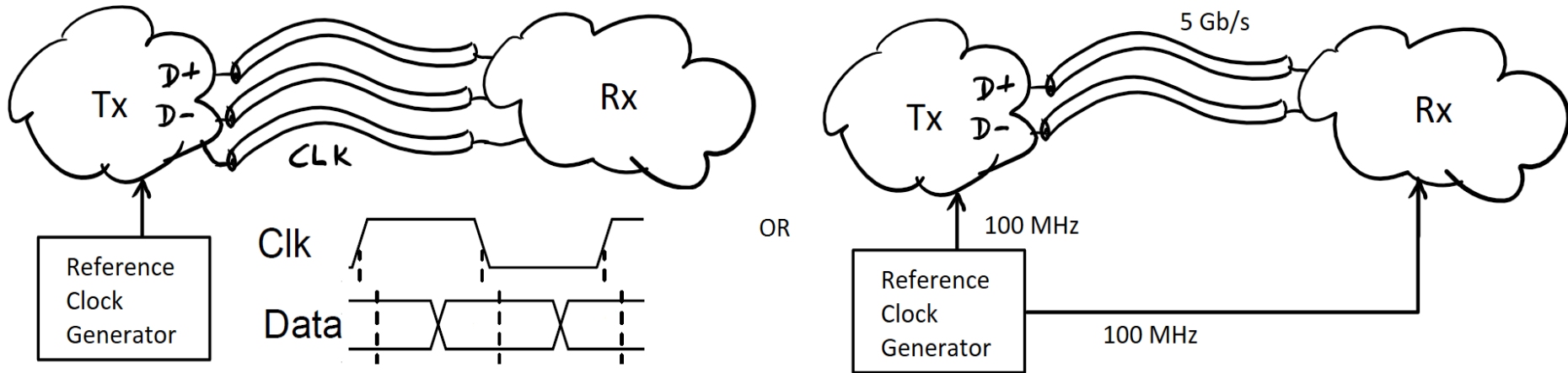


Synchronous System



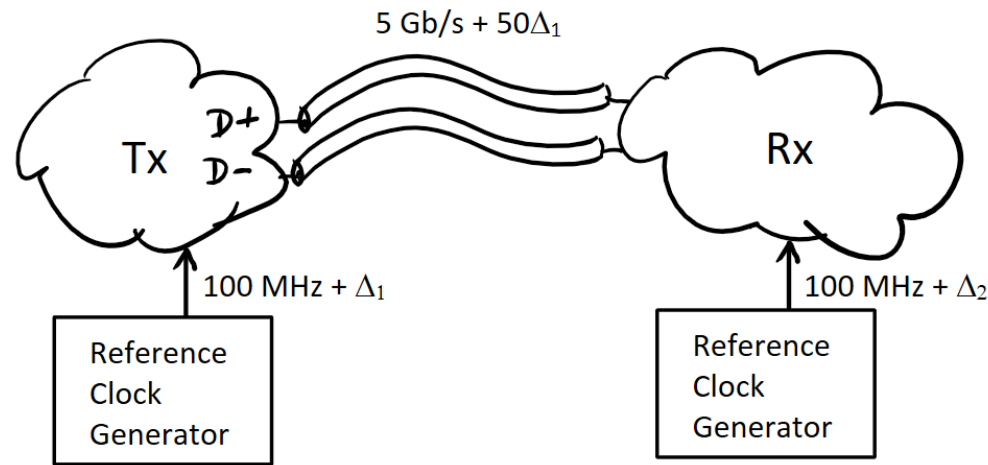
- Clock and data are aligned in phase
- Possible for low speed systems, such as SPI bus, low rate serial or parallel interfaces
 - Good for configuring the ICs (easy to design and reliable)
- Not possible to use them for high speeds, and unattractive for long interconnects (as an additional wire for the clock is required).

Mesochronous System



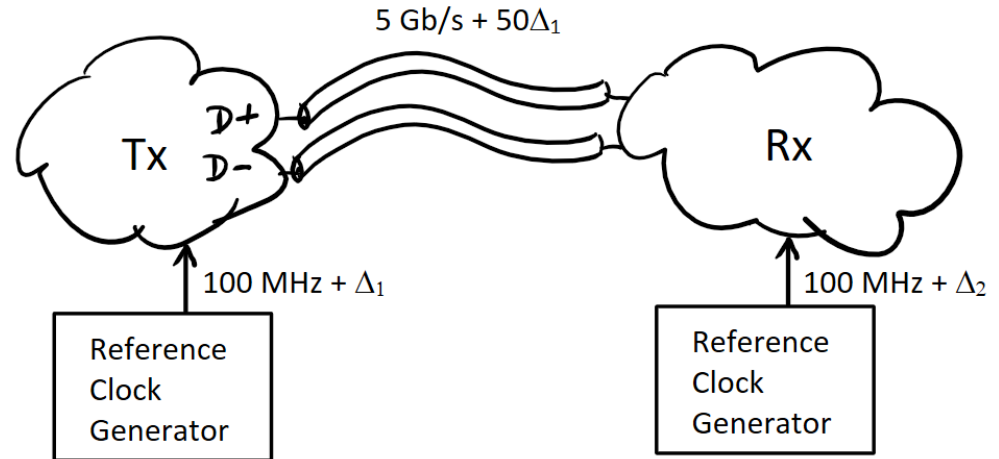
- Clock and data are synchronous with same clock frequency, may have an unknown phase offset
- Clock going to the receiver may be the full rate or an N^{th} rate clock.
- Rx may require a PLL to generate the desired frequency clock
- Generally applicable to on board chip-to-chip links, FPGAs with high-speed IO interfaces

Plesiochronous System



- The reference clocks for the two systems may have a small frequency offset (although they're meant to be the same).
 - Even a 1-Hz offset causes the data clock phase to shift by 2π in 1 second.
- Clock embedded clock must be recovered from the incoming signal
- To adjust for frequency difference between the Tx and Rx clock domains.
 - Rx has to employ a FIFO buffer
 - Tx has to send a few redundant words that can be dropped by the Rx if the it clock frequency is slower.

Plesiochronous System: Rate Mismatch Example

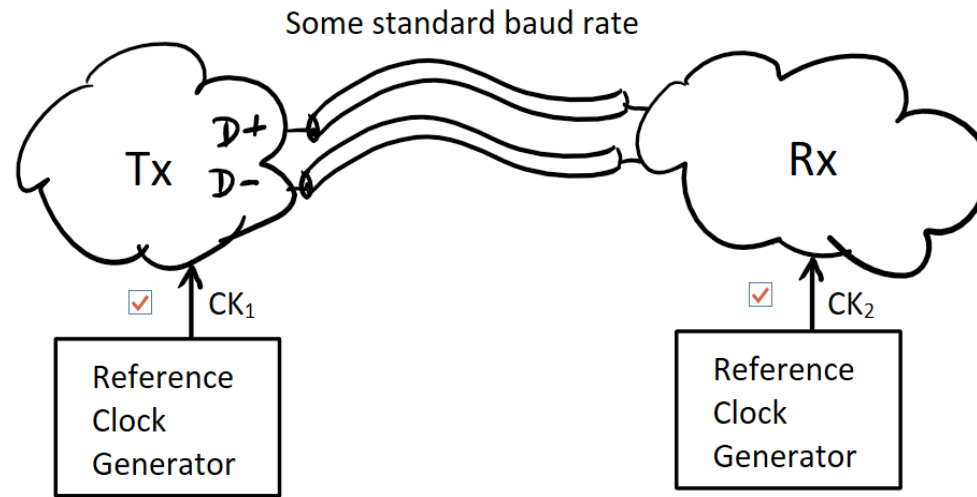


In the above figure, if $\Delta_1 = +1 \text{ kHz}$ and $\Delta_2 = -2 \text{ kHz}$, and a packet of transmitted bits contains 1000 bits. How many redundant data packets should the transmitter be sending per second to ensure that the Rx doesn't have to drop any useful data?

Assume that the transmitter transmits at "5 Gbps" and the receiver internal circuitry can receive at "5 Gbps" according to their own 100 MHz clock references, respectively.

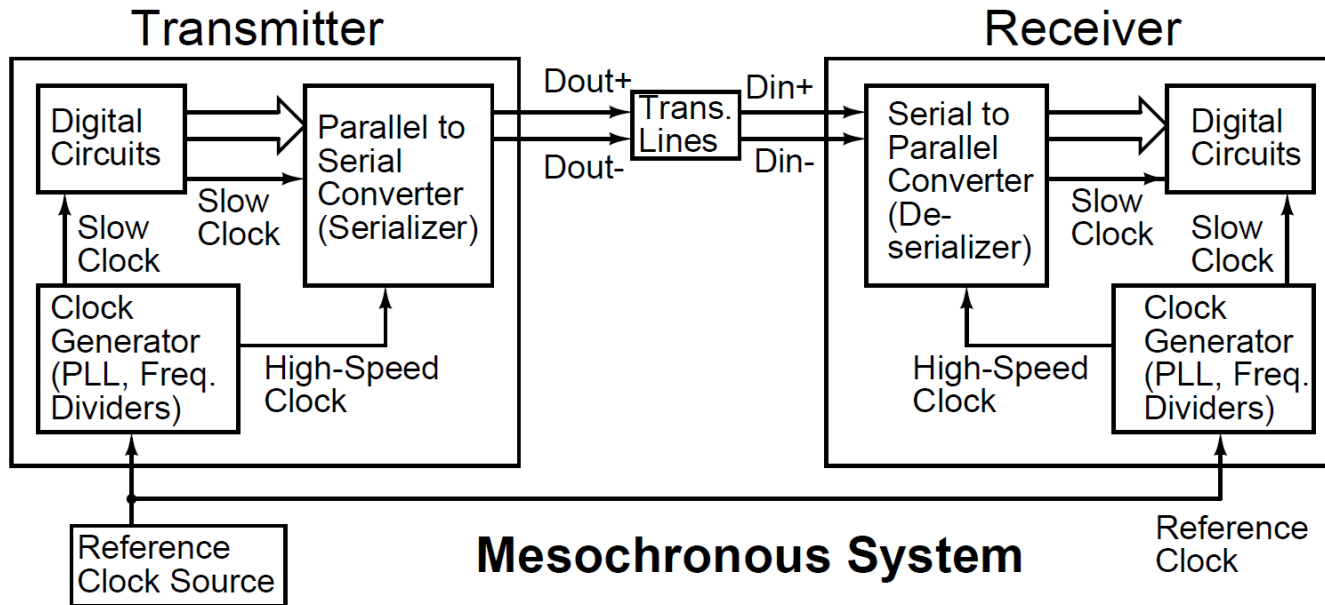
Answer: 150 packets/second (of 1000 bits each).

Asynchronous System



- The reference clocks for the two systems are independent
- Transfer of data is generally packed based and the data transfer rate is not very well defined
- For example, some standard baud rates used in UARTs in bps include 300, 1200, 2400, 4800, 9600, 14400, 19200
 - Transitions between stop/start bits are used for “synchronizing” at the byte level (for a given approximate baud rate)
- Sometimes handshake signals used for data transmission
- Greatly limits the transfer rates.

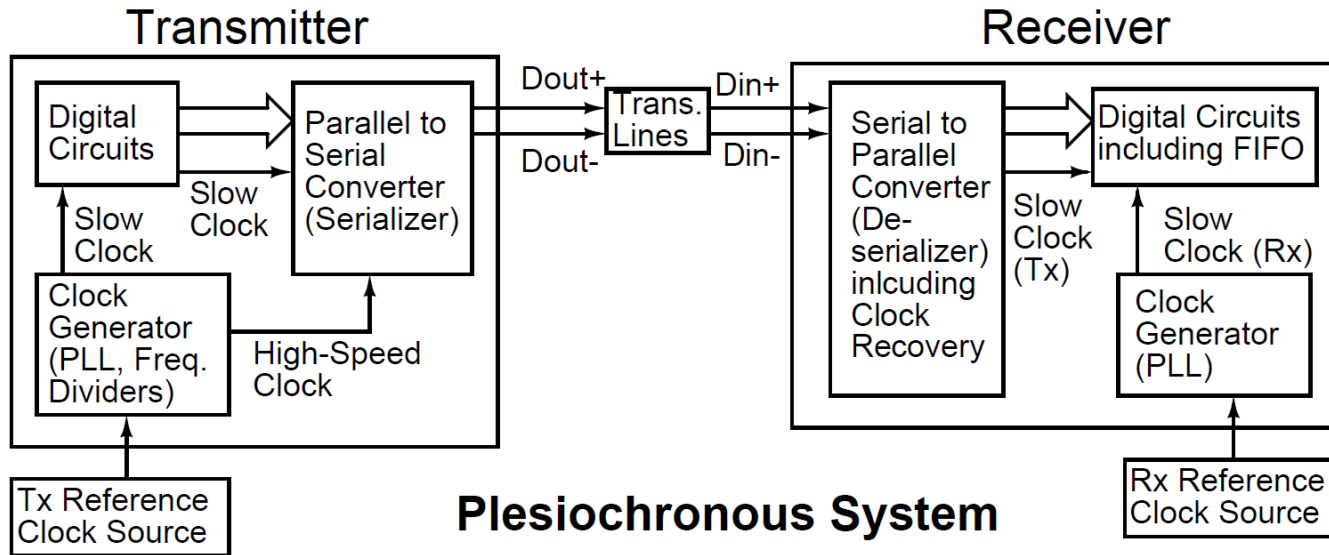
SerDes for Mesochronous Systems



Mesochronous System

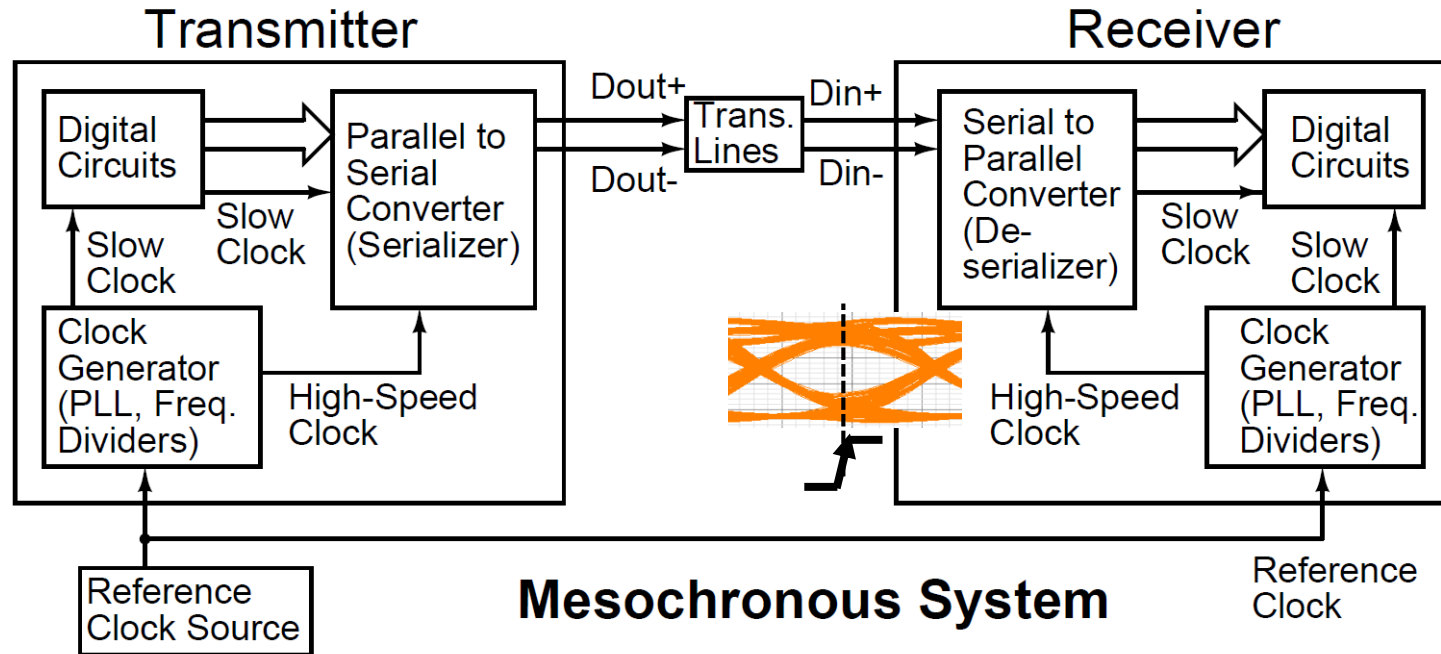
- High speed serial links comprise SerDes (Serializer-Deserializer)
- Mostly used in Mesochronous or Plesiochronous Systems
- May include pre-emphasis filter at Tx output and feed-forward/decision-feedback equalizer at Rx input
- No need to clock recovery and FIFO for rate matching in Mesochronous systems

SerDes for Plesiochronous Systems



- Receiver requires a clock recovery circuit to recover the clock embedded in the data signal
- Two system clocks: Tx and Rx rates have a mismatch
- FIFO buffer has to be added in the digital domain to take care of the rate mismatch
 - Some redundant packets are added at the Tx, which may be dropped by FIFO if Tx clock is faster than Rx clock

De-Serializers: Phase Alignment



Deserializers require alignment of clock-phase wrt the data signal to sample the data correctly

- Mesochronous systems: Require a Delay-Locked-Loop (DLL) for clock phase alignment
- Synchronous systems: Require a Phased-Locked-Loop (PLL) for clock recovery and phase alignment



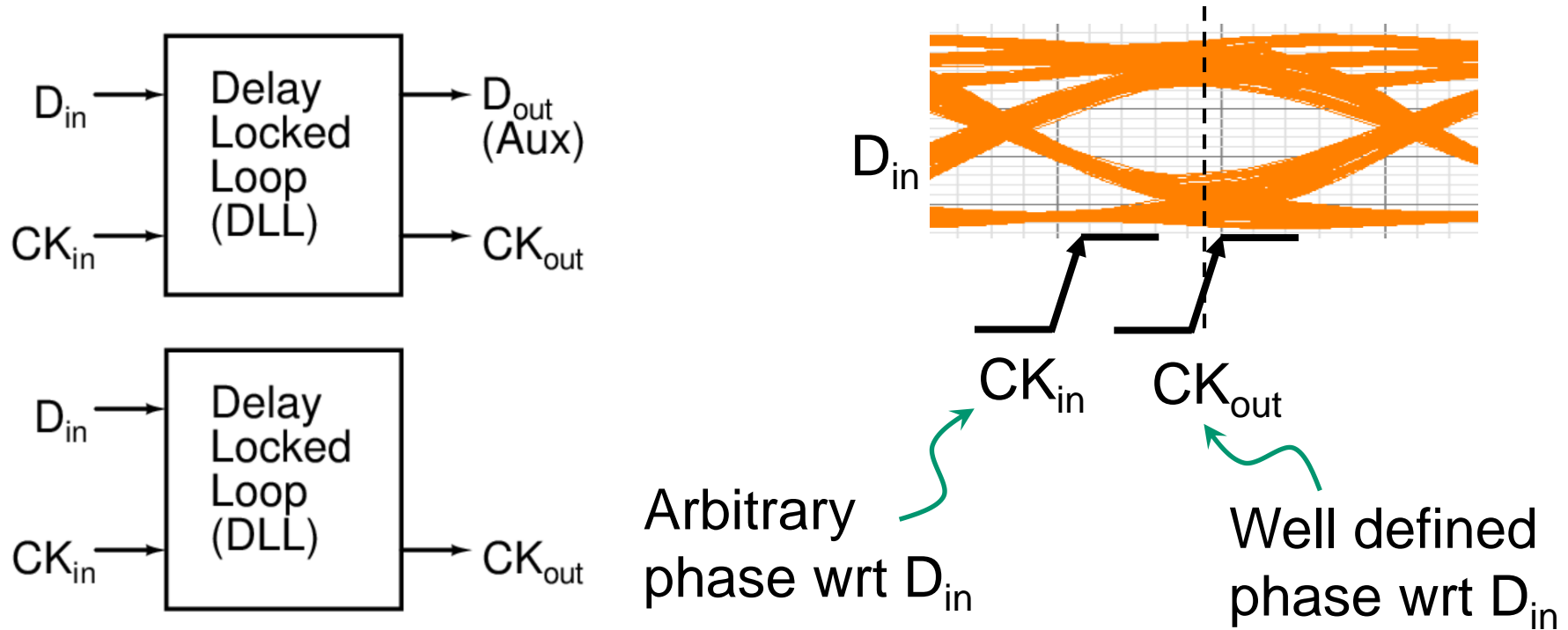
DLLs and PLLs

A Delay Locked Loop adds the desired phase shift (or delay) to the periodic input signal using a feedback technique.

A Phase Locked Loop generates a periodic signal with a well-defined phase with reference to a periodic or clock synchronous input signal.

A PLL has an inbuilt tunable oscillator, whereas a DLL has tunable delay element(s) but NO oscillator!

Delay Locked Loops



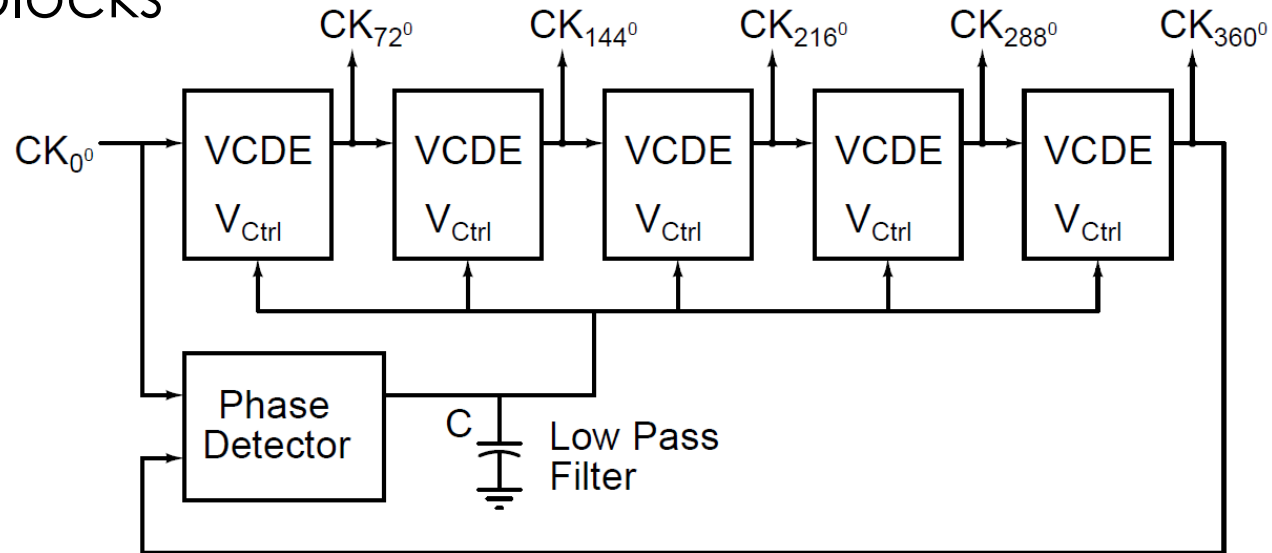
A delay locked loop (DLL) aligns the input clock phase with respect to the input signal

- Typically provides an output clock that is a phase shifted version of the input clock

Delay Locked Loops

A delay locked loop can be used for consists of

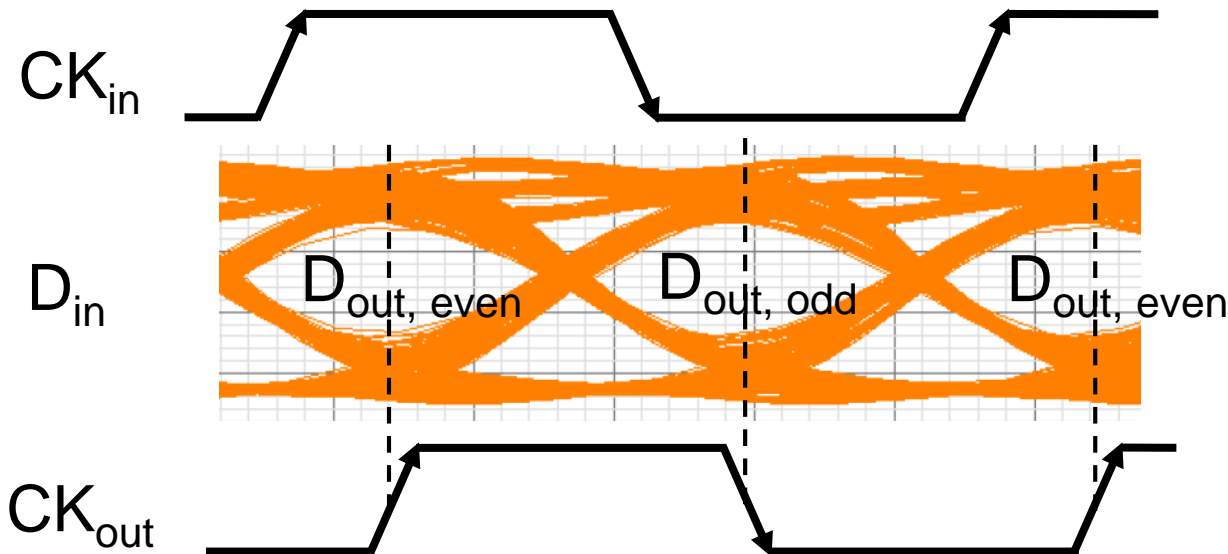
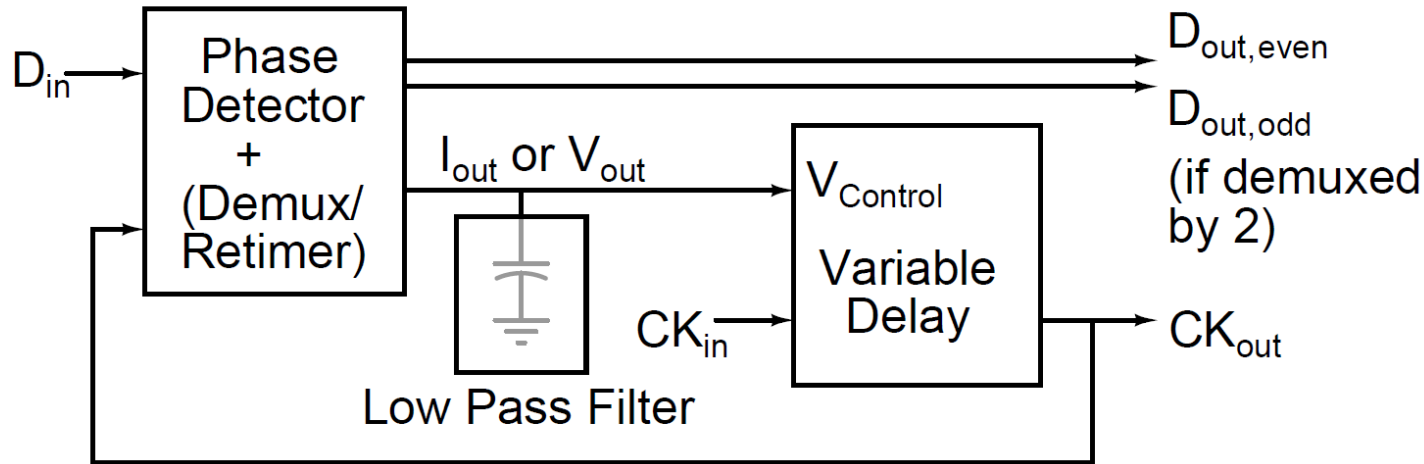
- Phase Detector
- Charge Pump (optional)
- Loop filter (or a low pass filter)
- Tunable delay blocks



Can use for

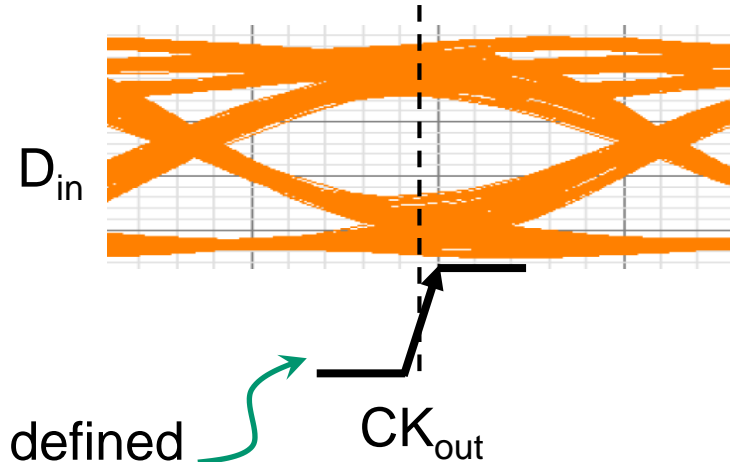
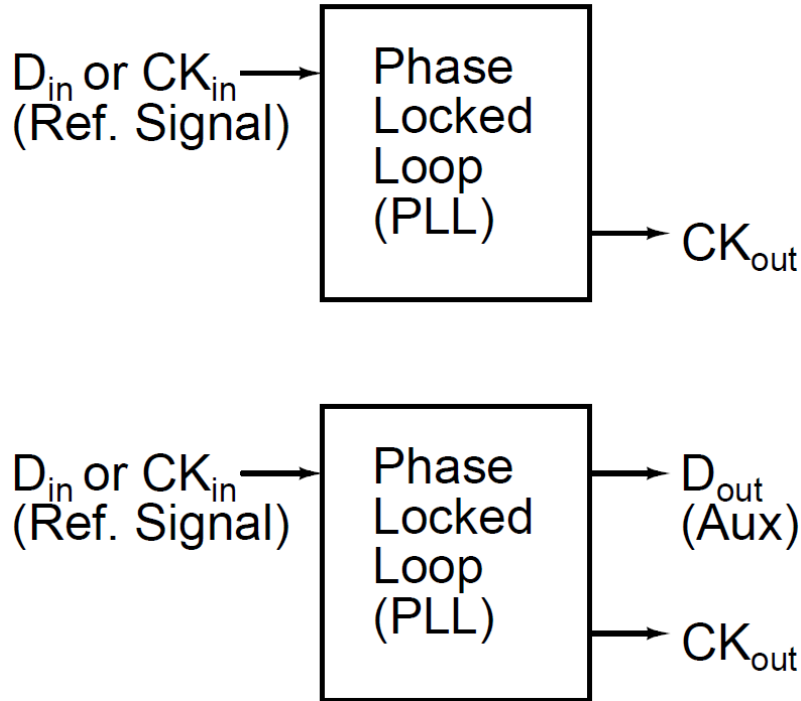
- clock alignment with data
- multi-phase clock generation

Delay Locked Loops: Half-Rate PD with 1:2 Demux



Half-rate: The input clock frequency is half the input data-rate

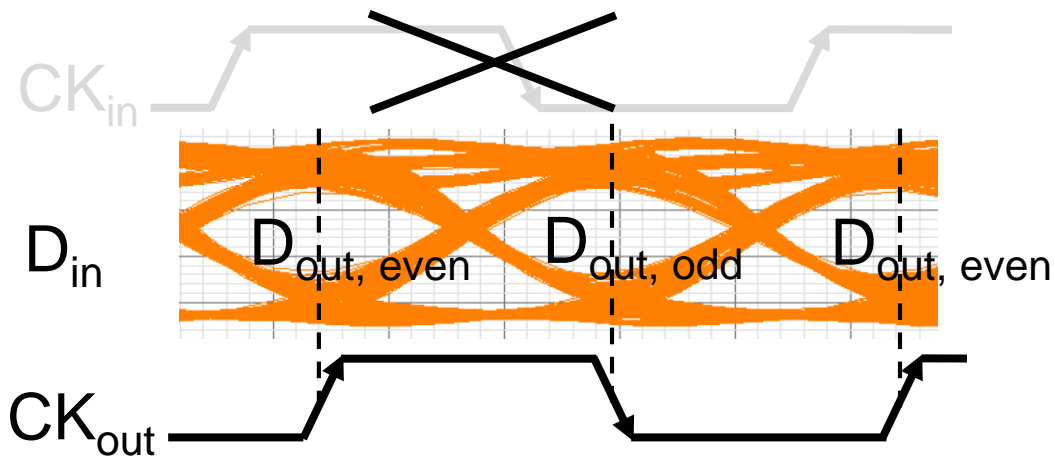
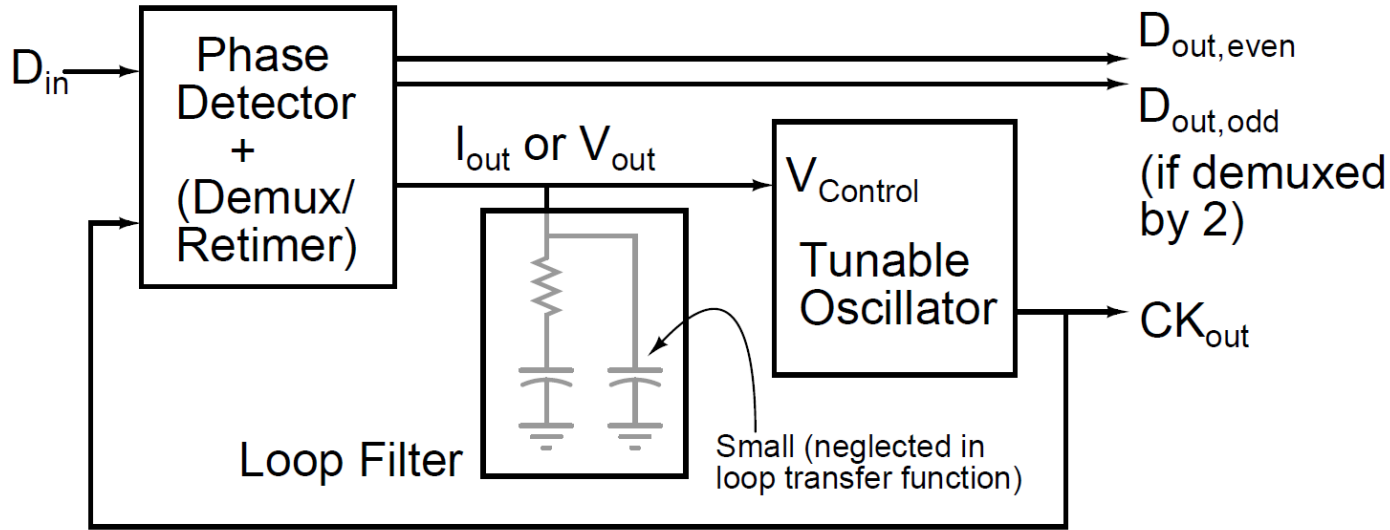
Phase Locked Loops



The phase locked loop (PLL) generates a clock and aligns its phase (and therefore frequency) w.r.t. the input signal phase

- Generates the clock on its own (separate external clock not required)

PLL Based CDR Example: Half-Rate PD and 1:2 Demux



The same half-rate PD can be used in a PLL configuration as well (for recovering data in a plesiochronous system)

DLLs vs. PLLs



| Property | DLLs | PLLs |
|-----------------------------|--|--|
| Tunable block | Delay(s) | Oscillator |
| Loop transfer function | Has one pole at zero frequency | Has at least two poles at zero frequency |
| Loop filter | One pole is sufficient | Has a pole; should also have a zero for stability |
| Application in Serial Links | Data Recovery (Mesochronous Systems) | Clock and Data Recovery (Plesiochronous Systems) |
| Stability | Unconditionally stable | Stability NOT guaranteed (requires careful analysis) |
| Loop Bandwidth | High (fast locking and tracking) | Low (slow locking and tracking) |
| Jitter response | Transparent to ref. clock jitter (ref clock and data jitters track each other) | Low pass filter input jitter; High pass filter VCO jitter. |

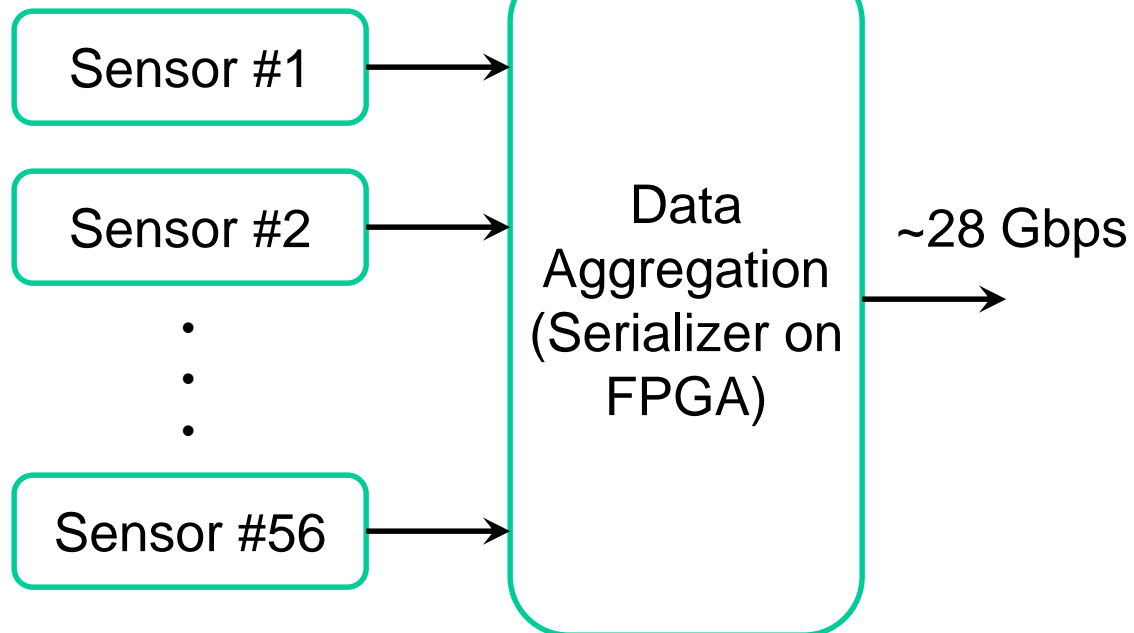
Ideal system for HEP experiments



Distribute and use common reference clock
(Mesochronous System)

- Need DLLs instead of PLLs: Faster settling, no need for large FIFO buffers to adjust for rate mismatch

Say ~500 Mbps each

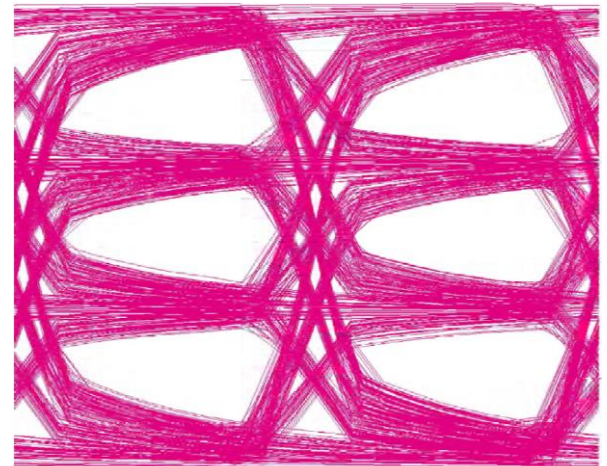


Need for large buffers and associated delays is obviated if all sensors use the common reference clock

Other Aspects of Serial Links



- Need for line-coding schemes (such as 8b-10b or 64b-66b) to aid clock recovery and framing bits (depending on protocols)
- Equalization required to compensate for low-pass nature of the channel
- PAM-4 for higher bit-rates where channel is bandwidth limited and not SNR limited



Eye diagram for PAM-4



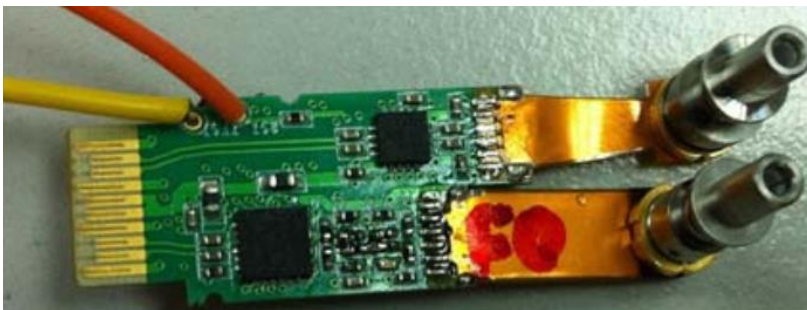
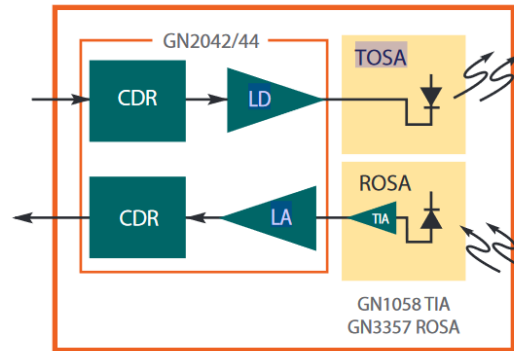
Optical Interfaces

Optical Interfaces

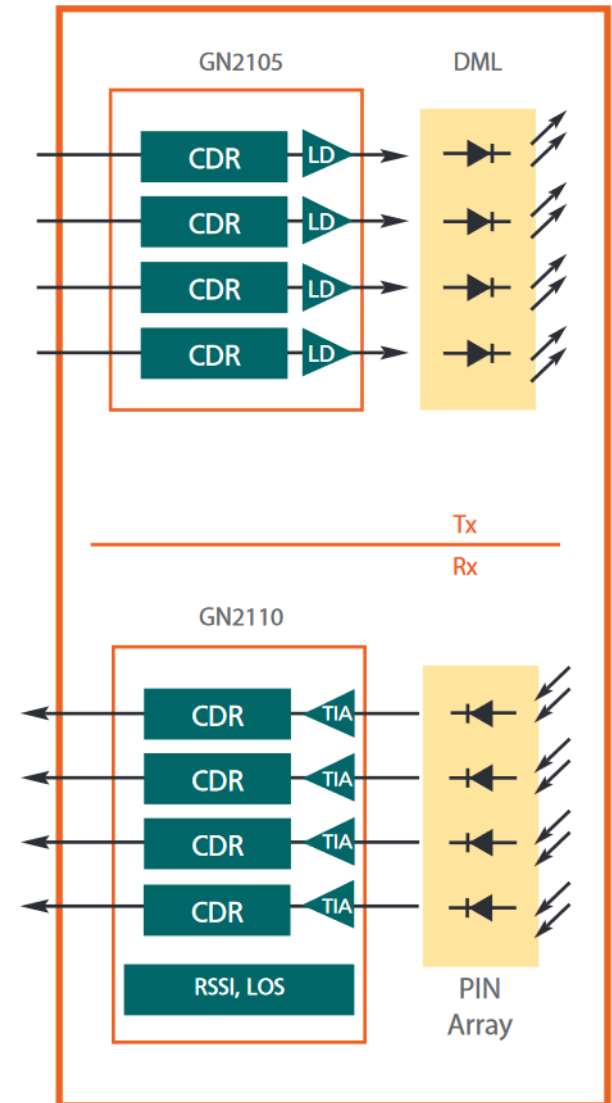
- Optical channels provide much higher capacity, longer reach
- Various modules with small form factors are available
- Direct modulation and detection of lasers

Examples

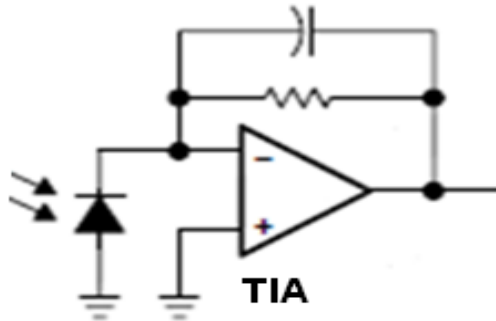
XFP, SFP+: 10G Retimed



QSFP28/CFPx: 100G PSM4/CWDM4



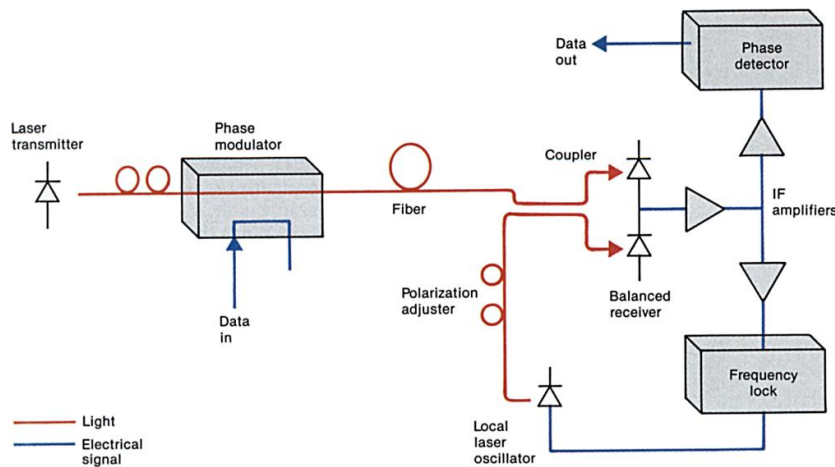
Optical Links: Historical Perspective



Direct Modulation Direct Detection

- Common in legacy links
- Still power hungry, spectrally inefficient.
- Wavelength division multiplexing can also be used
- 100GbE short distance links use 4 lambdas or PAM-4/PAM-8 for spectral efficiency, lower power

Optical Links: Coherent Systems



II. Coherent optical detection experiments

| Organization | Modulation scheme | Transmission rate, Mb/s | Sensitivity, photons per bit | Unamplified distance, km | Rate-distance product, Gb/s times kilometers | Semiconductor laser type |
|--|------------------------|-------------------------|------------------------------|--------------------------|--|--|
| British Telecom Research Laboratory Ipswich, England | Frequency-shift keying | 140 | 170 | 200 | 28 | Narrow line |
| Nippon Electric Co. Kawasaki, Japan | Frequency-shift keying | 280 | 550 | 204 | 57 | Broad line |
| AT&T Bell Laboratories Holmdel, N.J. | Phase-shift keying | 400 | 45 | 257 | 103 | Narrow line |
| Nippon Telegraph and Telephone, Tokyo | Frequency-shift keying | 400 | 195 | 290 | 116 | Narrow line |
| Bell Communications Research, Red Bank, N.J. | Frequency-shift keying | 560 | 1670 | 38 | 21 | Broad line |
| AT&T Bell Laboratories | Phase-shift keying | 1000 | 270 | 200 | 200 | Narrow line |
| Bell Communications Research | Frequency-shift keying | 1000 | 1500 | 100 | 100 | Broad-line transmitter Narrow-line receiver |
| AT&T Bell Laboratories | Space shift | 2000 | 1200 | 150 | 300 | Narrow line |

[2] Coherent optical transmission experiments at Bell Laboratories employ phase modulation to reach 2 Gb/s, the highest bit rate recorded for coherent systems.

Richard A. Linke and Paul S. Henry
IEEE Spectrum, February 1987

Coherent Systems

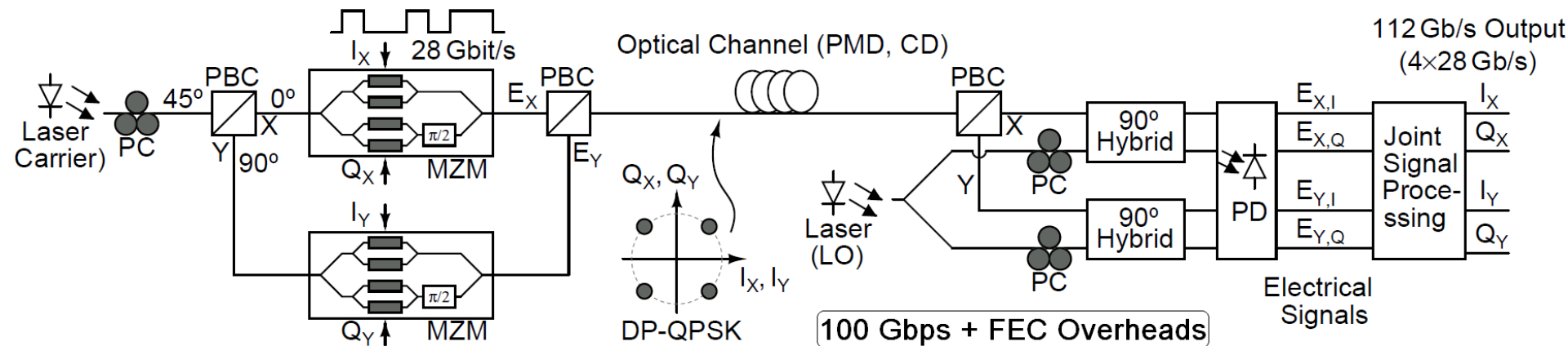
- Receiver needs reference phase to demodulation message
- Linear w.r.t optical field (direct detection is a non-linear process, difficult to invert the channel)
- Serious research started in early 1980s

Abandoned: Difficult; DMDD with WDM was sufficient at that time!

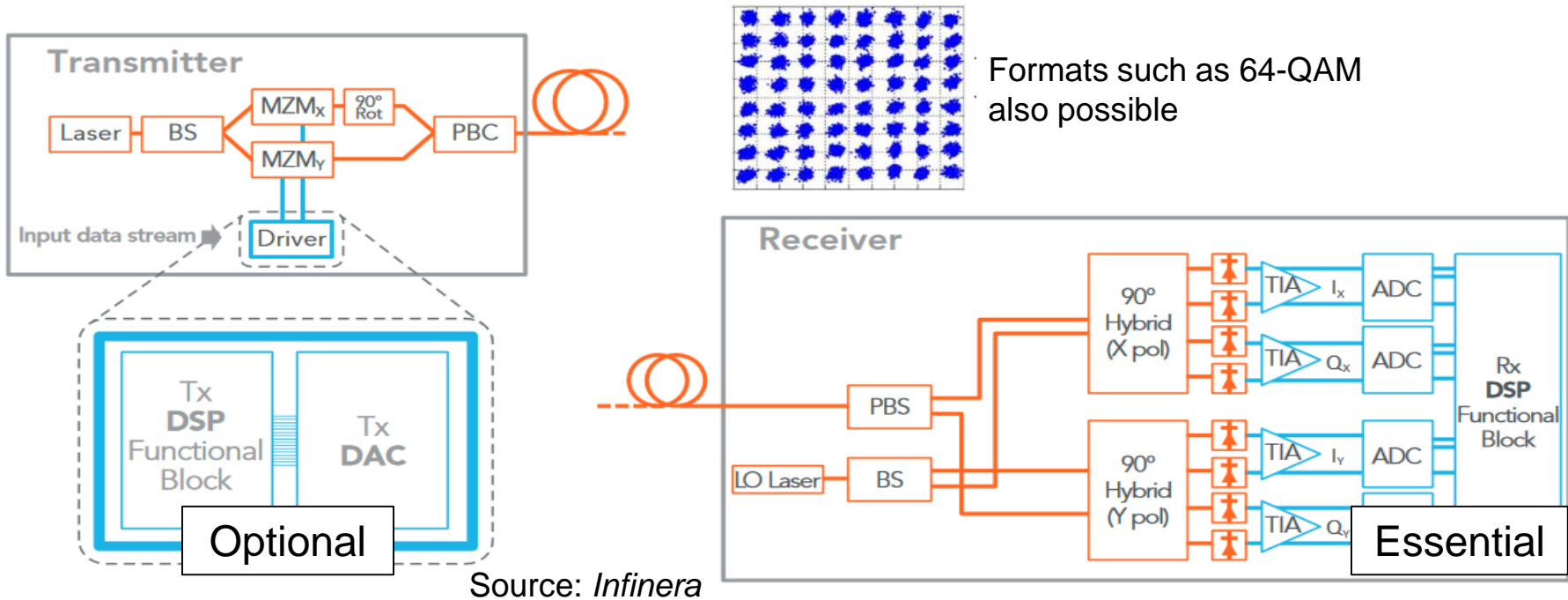
Progress in the last 10 Years

High-speed coherent research picked up again in ~2006

- Coherent 100Gbps DP-QPSK products easily available
- Record breaking results achieved (Pbits/s with DWDM and dense constellation such as 128-QAM)
- Key Enabler: Electronics (DSOs, AWGs, ADCs and DSPs)!
- Focus now on short distance links for data centers!



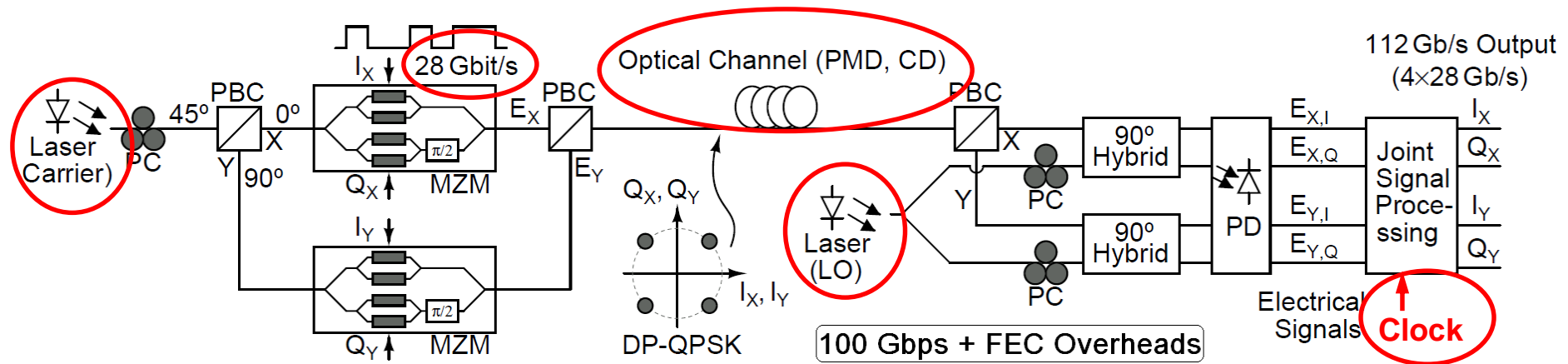
Coherent Optical Links



Coherent links: 100-Gbps / λ and beyond

- Both amplitude/phase of optical field modulated at transmitter
- Coherent detection at receiver (optical field is mixed with a local oscillator) to measure complex optical field (both amplitude and phase)
- Polarization multiplexing possible (due to linearity)
- Currently used only for long haul

Challenges in Coherent Transmission



- Challenges include
 - Optical channel impairments:
 - Dynamically varying Polarization Mode Dispersion (PMD) & Polarization rotation
 - Chromatic Dispersion (CD)
 - Optical non-linearity (channel-to-channel cross-talk)
 - Laser phase noise and frequency offset
 - Electronic clock offset

Entire burden is on electronics to recover data in the presence of these effects



Recent Focus: Data Centers

News & Analysis

Facebook Likes 100G at \$1/G

Rick Merritt

8/28/2015 11:50 AM EDT
2 comments

Tweet 11 Share 78 G+ 5

Web giant plans 2016 Ethernet upgrade

SANTA CLARA, Calif. – Facebook helped define a low cost 100G transceiver it expects to start deploying next year. The move at data centers are driving design in computers and networking.

Facebook specified a 100G transceiver using single-mode fiber. **To hit the lower costs it relaxed distance requirements to 500 meters down from 2km and eased specs on operating temperature and product lifetime.**

Inphi Debuts 100G DWDM Solution for 80km Data Center Interconnects

ColorZ™ First to Enable 4Tb/s Bandwidth in QSFP28 Form Factor while Dramatically Lowering Power and Cost

SANTA CLARA, Calif., March 22, 2016- Inphi Corporation (NYSE: INPH) provider of high-speed connectivity solutions, today announced the release of its "ColorZ" reference design, the industry's first Silicon Photonics 100G PAM4 platform solution for 80km DWDM Data Center Interconnect (DCI) in QSFP28 form factor.

Utilizing advanced Pulse Amplitude Modulation (PAM4) signaling, ColorZ delivers up to 4Tb/s of bandwidth over a single fiber and allows multiple data centers located up to 80km of each other to be connected and act like a single data center. ColorZ provides

OFC 2016

OFC
The Optical Networking and Communication Conference & Exhibition
Technical Conference: 11 - 15 March 2018
Exhibition: 13 - 15 March 2018
San Jose, California, USA

ABOUT PROGRAM & SPEAKERS EXHIBIT HALL & FLOOR PROGRAMS SUBMIT PAPERS

HOME > PROGRAM & SPEAKERS > 2017 PANELS

Direct vs. Coherent Detection for Metro-DCI

Access Technical Papers

- 2017 Plenary
- 2017 Invited Speakers
- 2017 Tutorial Speakers
- 2017 Data Center Summit
- 2017 Demonstrations and Events

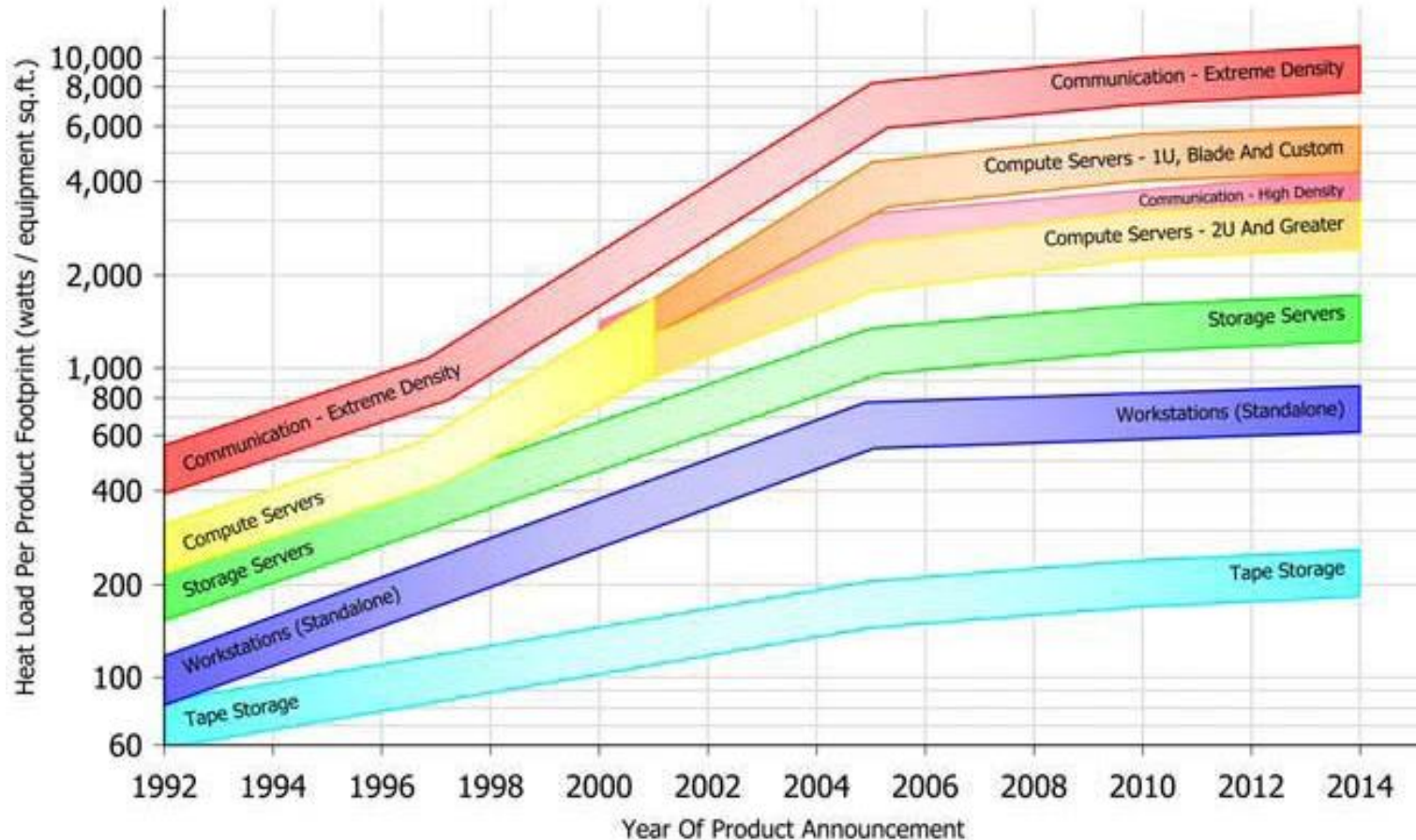
Tuesday, March 21, 2017
4:30 PM - 6:30 PM
Event type: Panel
Room number: 402AB
Organizer:
Robert Griffin, Oclaro, UK; Tom Issenhuth, Microsoft, USA; Sorin Tibuleac, Adva Optical, USA

OFC 2017 Panel

High-capacity long haul links have been there for a while. Focus now shifting to short distance Tbps optical interconnects



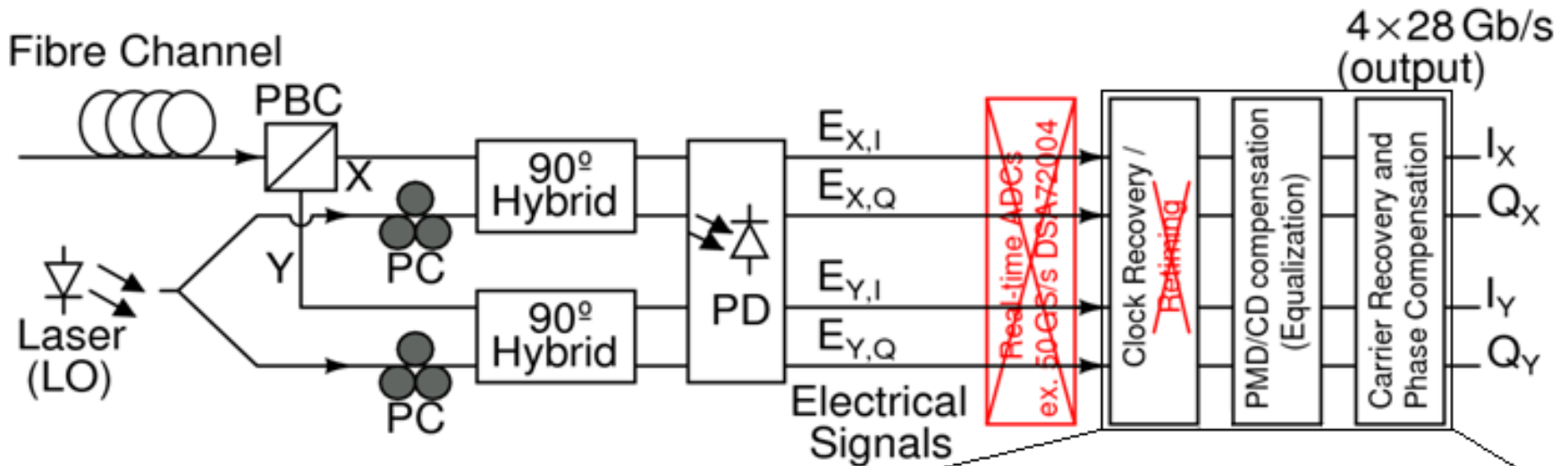
Power Consumption in Data Centers



C. Belady, "Data Center Power Projections to 2014," *ITHERM* 2006.

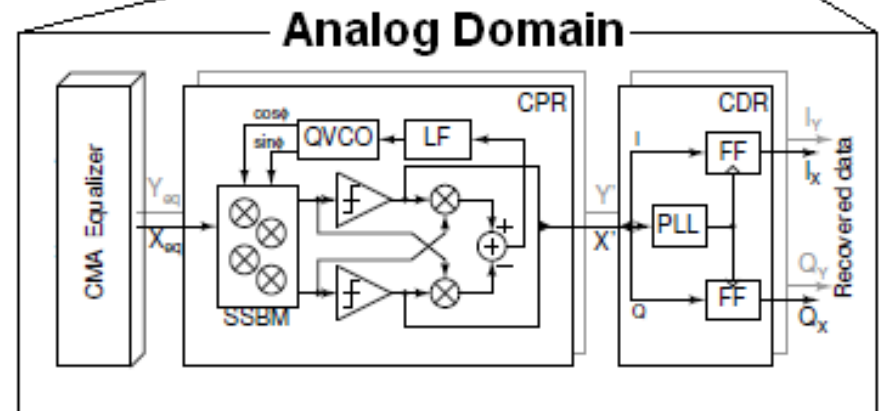
Power and cooling costs already higher than the cost of IT equipment in data centers

Coherent Rx: Analog Domain Signal Processing



Analog processing well suited for this application

- Power efficient at these speeds
- Much faster feedback control possible: easy compensation of dynamic changes (such as carrier frequency offsets)



First Demonstration of an All Analog Adaptive Equalizer for Coherent DP-QPSK Links

Nandakumar Nambath¹, Mehul Anghan¹, Nandish Thaker¹, Rakesh Ashok¹,
Rashmi Kamran¹, Arvind Kumar Mishra², and Shalabh Gupta¹

¹Department of Electrical Engineering, IIT Bombay, Mumbai – 400076, India

²Sterlite Technologies Ltd., Aurangabad – 431136, India

npnandakumar@iitb.ac.in

OFC 2017

Abstract: For the first time, an all analog CMA equalizer for DP-QPSK transmission systems has been demonstrated. The experiment with an 8-Gb/s link shows promise for low power analog processing based receivers for short-reach DP-QPSK links.

OCIS codes: 060.0060, 060.1660, 060.2360

1. Introduction

Internet traffic through data centers is anticipated to reach ~900 exabytes per month by 2019 globally [1]. Data traffic of this magnitude would result in a massive amount of power consumption in hundreds of billions of kWh and an

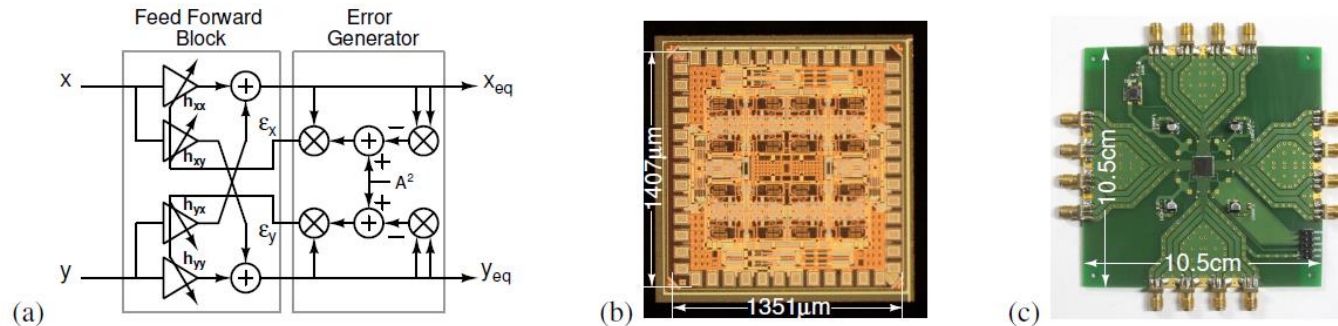


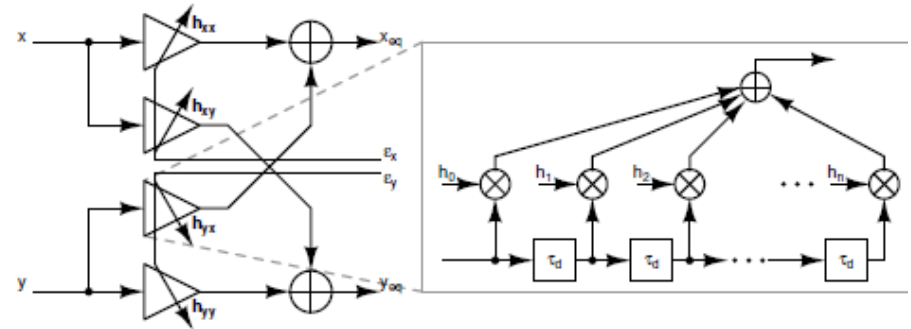
Fig. 1. All analog CMA equalizer: (a) architecture, (b) die micro-graph, and (c) PCB for low-speed validation.

Constant Modulus Algorithm

- Outputs are given by

$$x_{eq} = \mathbf{h}_{xx}^T \mathbf{x} + \mathbf{h}_{yx}^T \mathbf{y}$$

$$y_{eq} = \mathbf{h}_{xy}^T \mathbf{x} + \mathbf{h}_{yy}^T \mathbf{y}$$



Multidimensional adaptive equalizer

- The CMA update equations are given as:

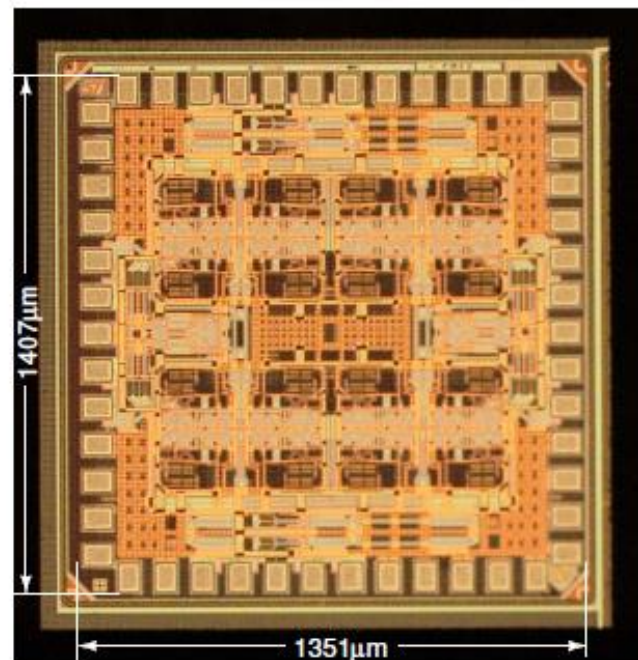
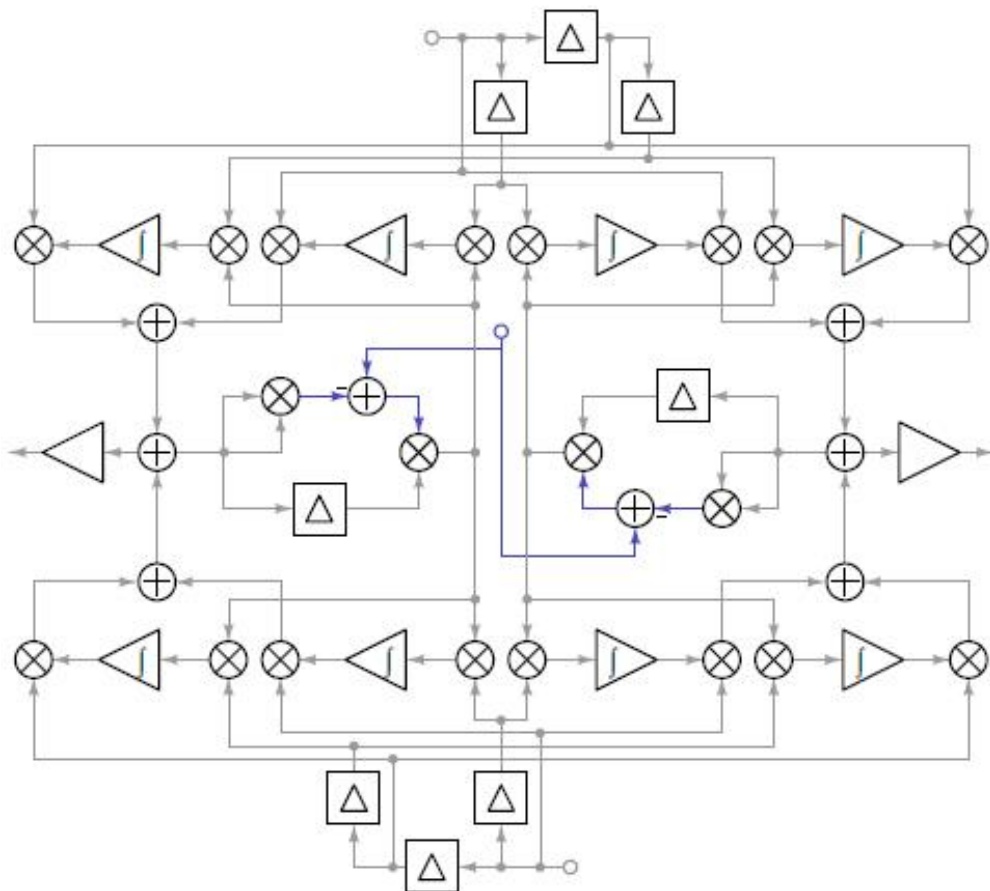
$$h_{xx,k}(t) = \beta \int_0^t x_{eq}(\tau) [1 - |x_{eq}(\tau)|^2] x(\tau - k\tau_d) d\tau$$

$$h_{yx,k}(t) = \beta \int_0^t x_{eq}(\tau) [1 - |x_{eq}(\tau)|^2] y(\tau - k\tau_d) d\tau$$

$$h_{xy,k}(t) = \beta \int_0^t y_{eq}(\tau) [1 - |y_{eq}(\tau)|^2] x(\tau - k\tau_d) d\tau$$

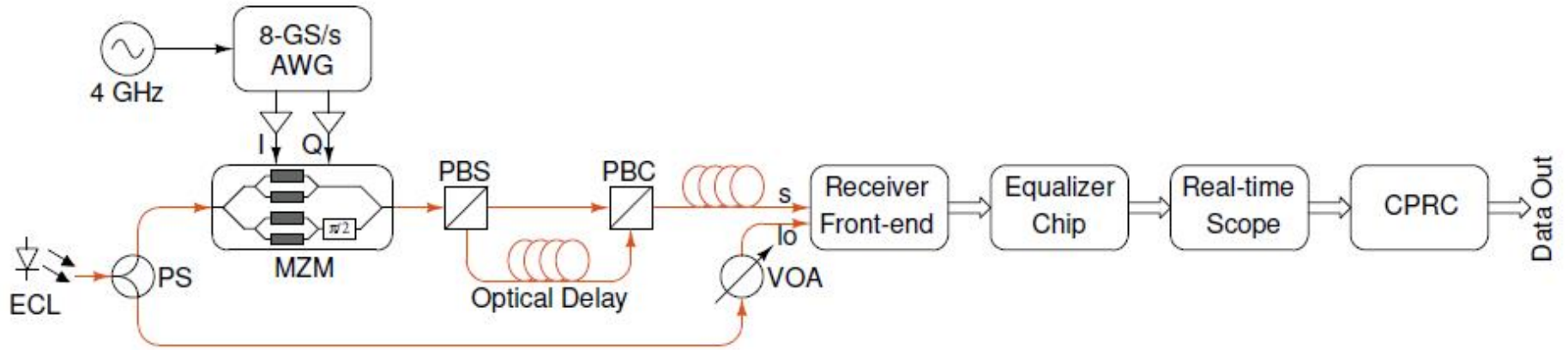
$$h_{yy,k}(t) = \beta \int_0^t y_{eq}(\tau) [1 - |y_{eq}(\tau)|^2] y(\tau - k\tau_d) d\tau$$

Equalizer Implementation in 130nm BiCMOS Technology

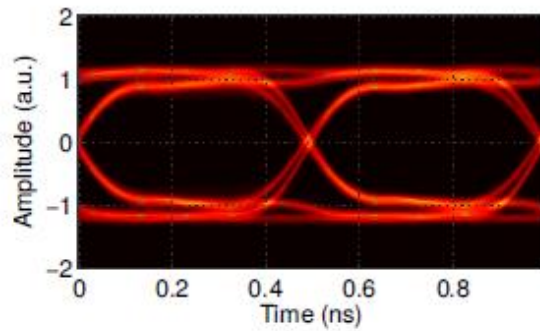


Chip area = 1.4mm1.35 mm, Power = 2.5W; Baud Rate: 25 Gbaud
Can be reduced further with advanced CMOS

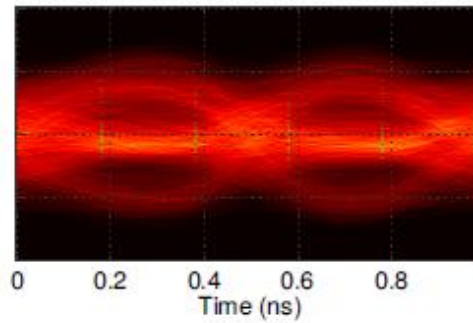
Chip Validation at 8-Gbps



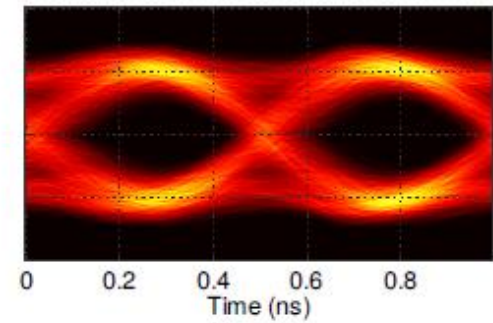
Experimental setup with 8-Gb/s optical system



Transmitted signal



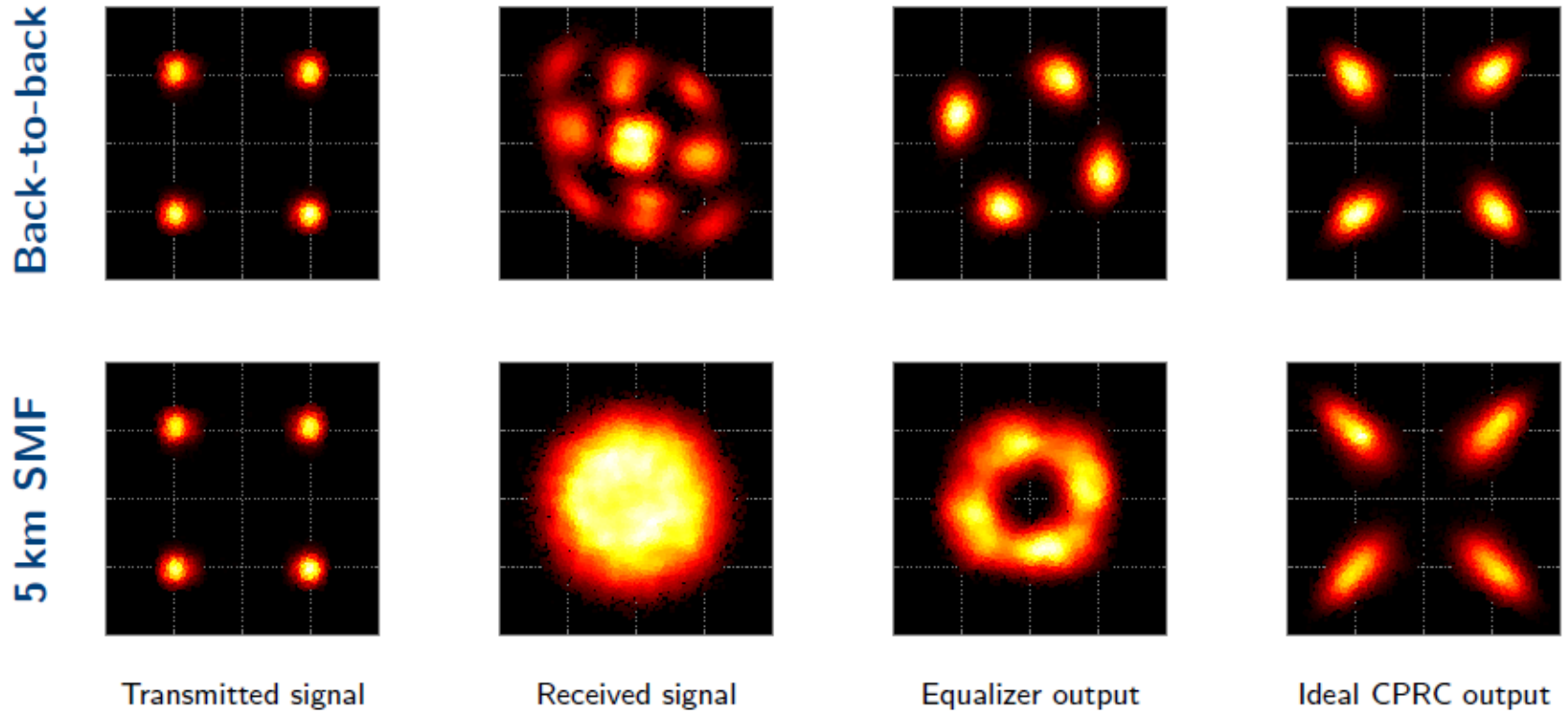
Received signal



Recovered signal

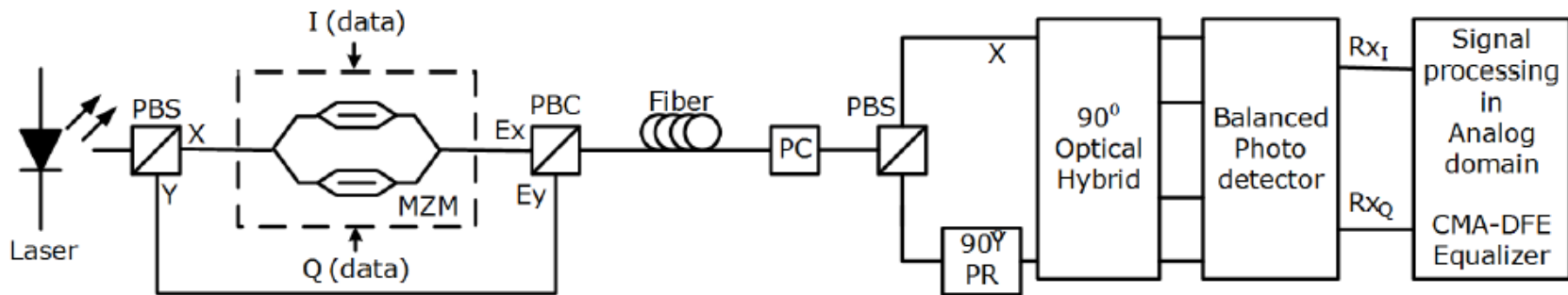
X polarization in-phase component eye-diagrams in the back-to-back link

Chip Validation at 8-Gbps



Error vector magnitudes: Tx - 12.5%, B2B - 19.8%, 5 km - 27%

Further simplification of the receiver: Self-homodyne (LO-less receiver)



Complex valued electric field representation for QPSK modulated signal and carrier signal are $E_x = A_S e^{(w_c t + \theta_m)}$ and $E_y = A_{LO} e^{w_c t}$ respectively

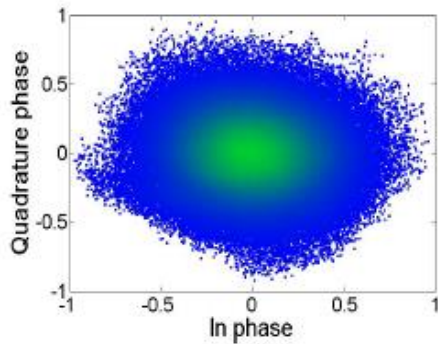
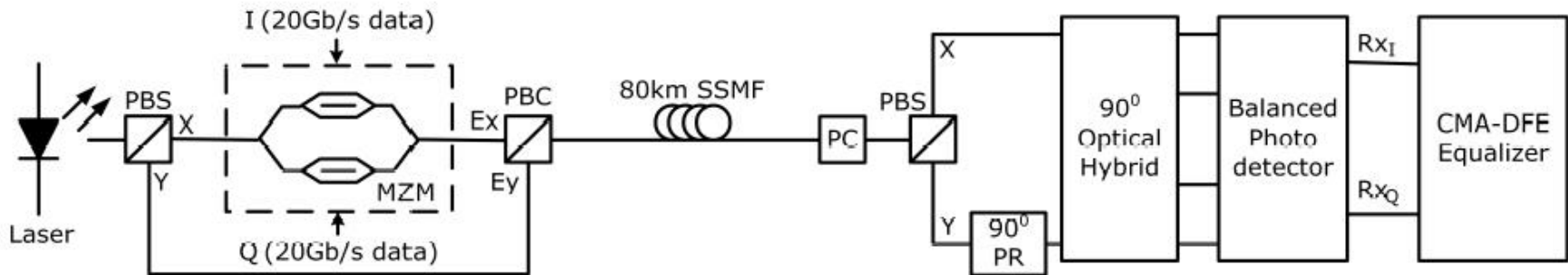
$$\begin{pmatrix} i_1 \\ i_2 \\ i_3 \\ i_4 \end{pmatrix} \approx \begin{pmatrix} [E_x + E_y][E_x + E_y]^* \\ [E_x - E_y][E_x - E_y]^* \\ [E_x + jE_y][E_x + jE_y]^* \\ [E_x - jE_y][E_x - jE_y]^* \end{pmatrix}$$

WOCC 2017

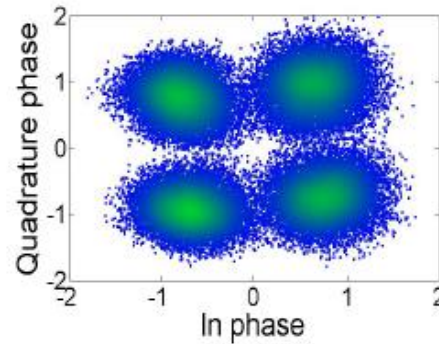
$$R_{X_I} = i_1 - i_2 = 4A_{LO}A_S \cos \theta_m$$

$$R_{X_Q} = i_3 - i_4 = 4A_{LO}A_S \sin \theta_m$$

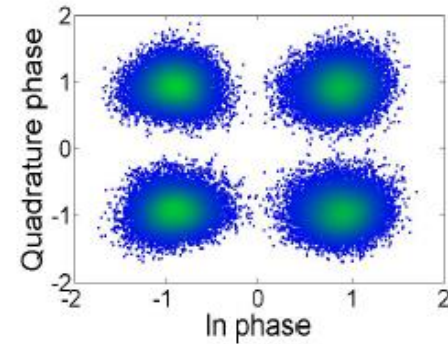
40 Gb/s SH-QPSK system with 80 km fibre



Received signal



CMA best output

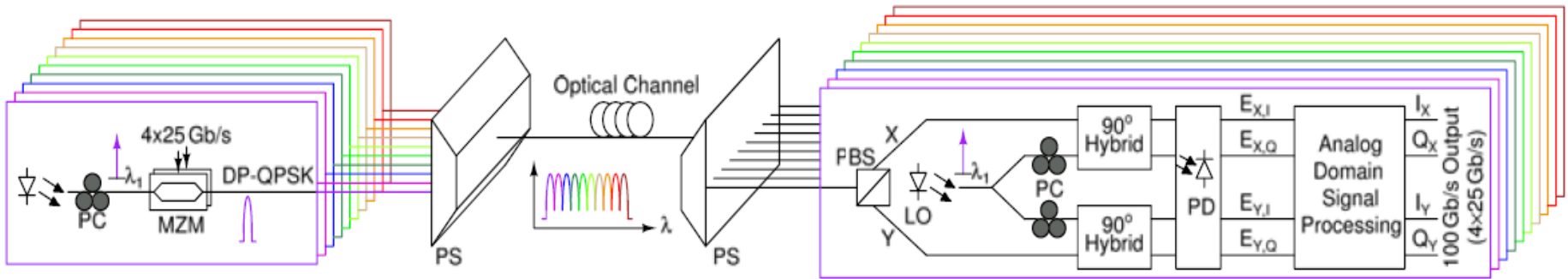


CMA-DFE output

CMA equalized signal with BER= 0.0059, $CMA_N=9$; CMA-DFE equalized signal with BER= 3.42×10^{-5} . $CMA_N=9$. $FF_N=1$. $FB_N=1$

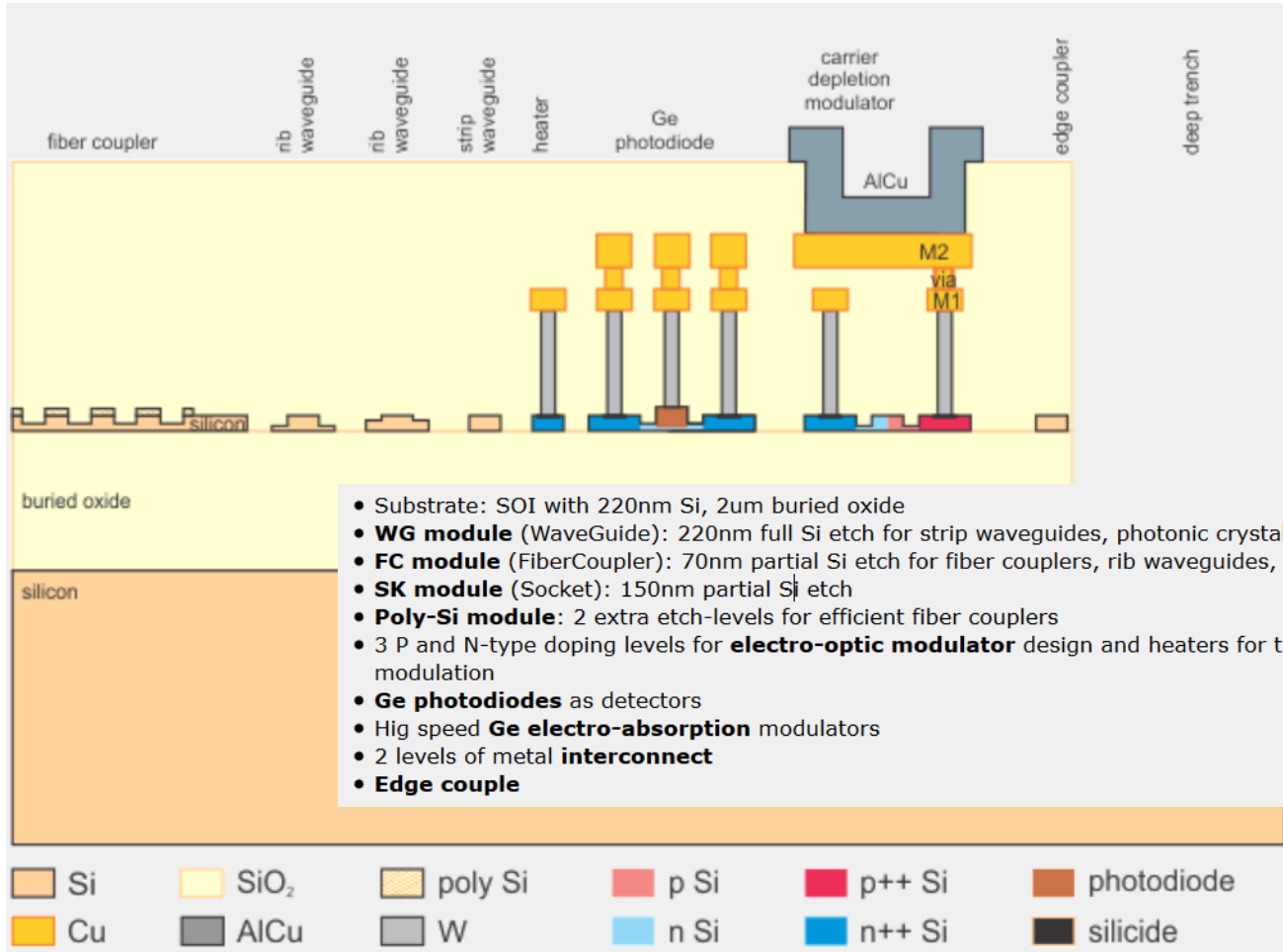
Requires CMA or CMA-DFE, that can again be implemented efficiently in analog!

Longer Term Vision: Ultra Low Power Tbps Optical Interconnects



- Use analog domain signal processing for low power consumption/form factor for short range coherent systems
- Simplified system using self-homodyne receiver.
- Use PICs along with analog processing for further reduction in power consumption and longer reach!

Further Integration Possible Using Photonic ICs



Photonic Integrated Circuits (using Silicon Photonics) being leveraged by IC designers for further integration



Acknowledgement

Would like to thank

- A large number of dedicated students and project associates who contributed to the project
- MeitY and DST (Govt. of India) for funding the projects
- Sterlite Technologies for giving access to their lab facilities for carrying out some of the experiments